



# Recent Developments in Transceiver SoC Design for Next Generation Optical Networks

Prof. C. Patrick Yue

Department of Electronics and Computer Engineering  
Hong Kong University of Science and Technology (HKUST)



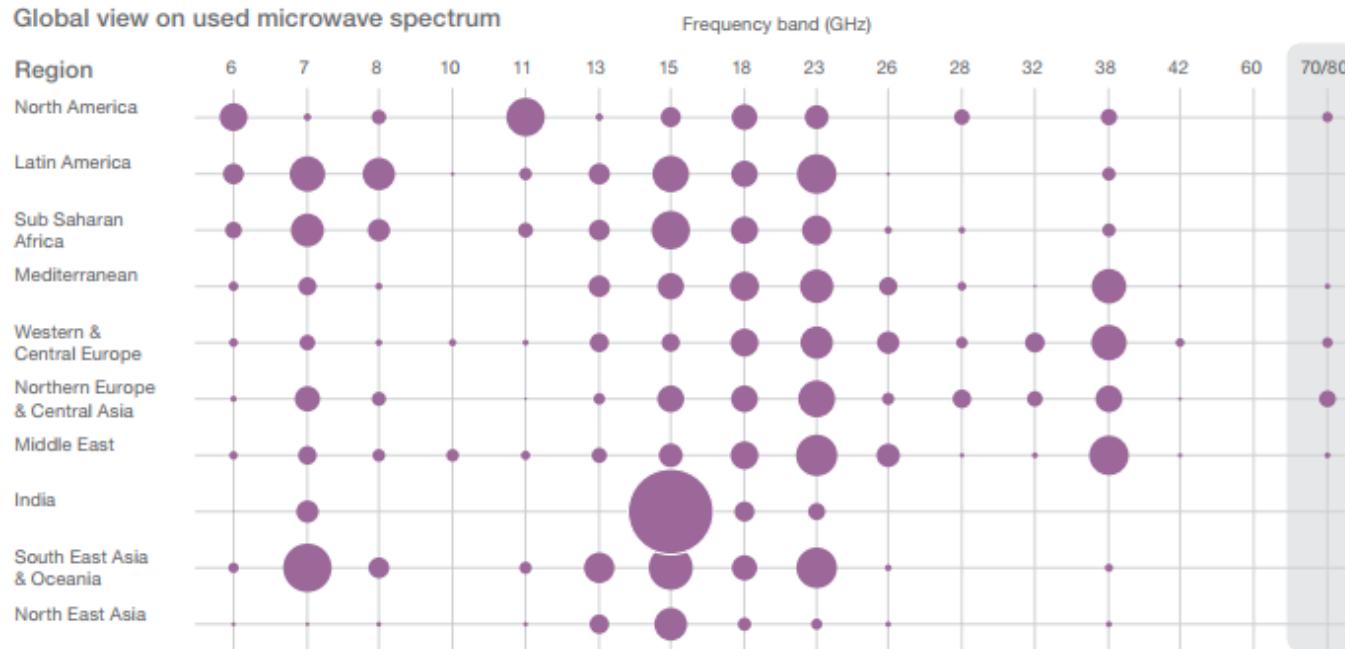


# Outline

- Motivation
- BoF circuit and system design
- Measured results
- Summary and future work

# Motivation

- Mobile data to grow >1000x over 2010 to 2020 [1]
- Expected 5G initial deployment in 2020 [1]

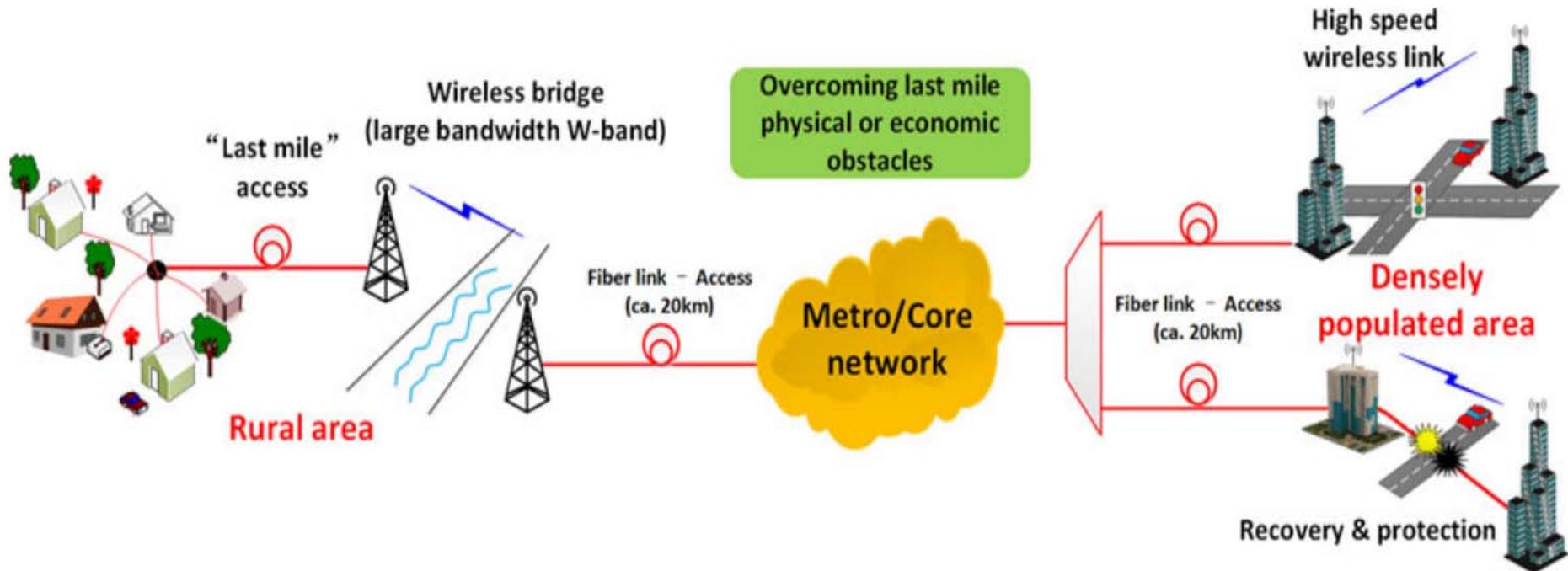


- New emerged mm-wave spectrum for 5G [2]
  - 26GHz, 28GHz, 32GHz, 38GHz, 70/80GHz...

[1] NTT DOCOMO 5G White Paper

[2] Ericsson Mobility Report

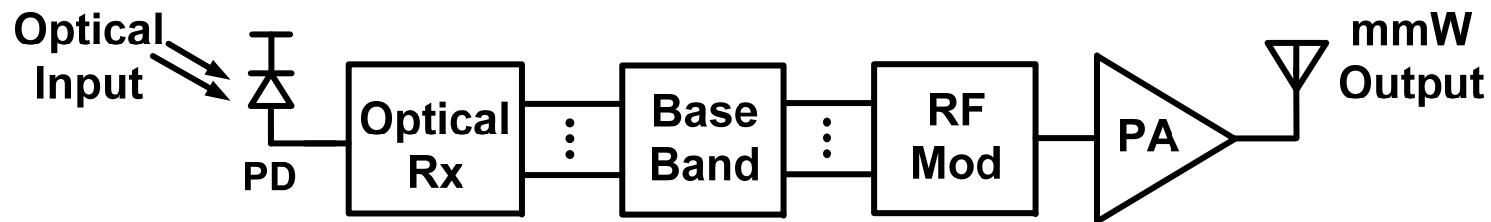
# Motivation



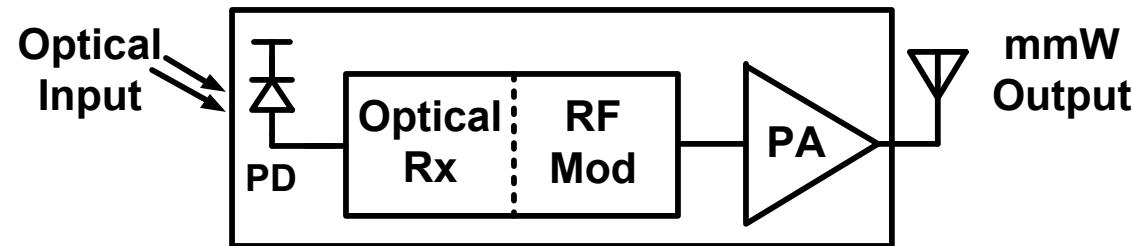
- Macro/pico cells widely used to extend coverage and data throughput capacity in 4G LTE and 5G cellular network
- Wired connection sometimes physically impractical or costly
- **Hybrid fiber-wireless network:** flexible and high performance solution for short-range backhaul links deployment

# Motivation

## Conventional Solution:



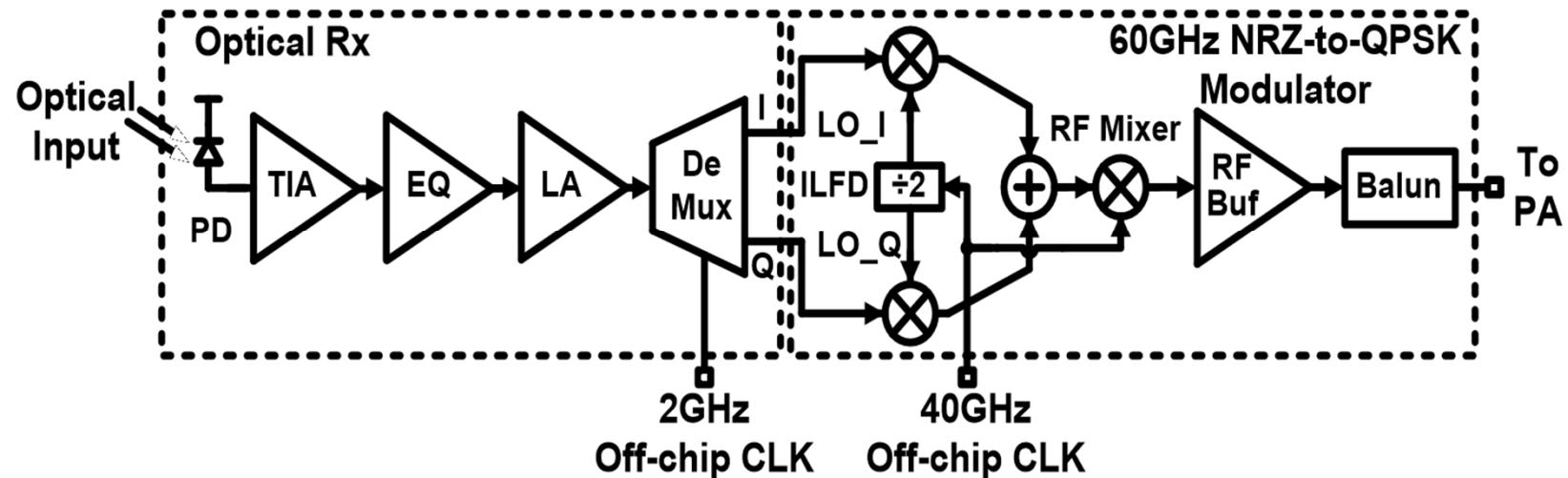
## Proposed Solution:



- Conventional fiber-wireless network composed of discrete O/E and RF components is costly and power hungry
- Redundant baseband digital processing introduces large overhead and latency

# Previous Work

## □ 60GHz 4Gb/s NRZ to QPSK Modulator [1]



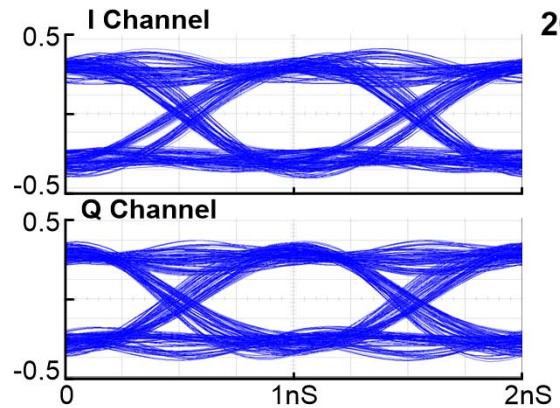
- Work at 60GHz
- No power amplifier integrated -> **external PA needed**
- No IQ calibration -> **may suffer from image tone**
- No on-chip demodulation -> **difficult to measure BER**

# Previous Work

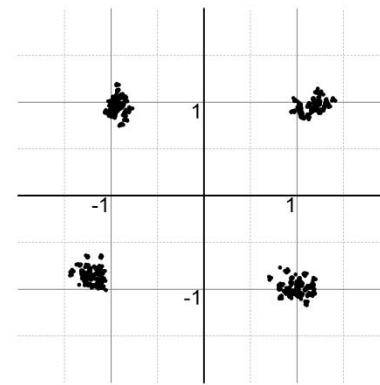
## □ 60GHz 4Gb/s NRZ to QPSK Modulator [1]

- Data rate limited to 4Gb/s
- Unbalance I/Q performance -> **May be IQ mismatch**

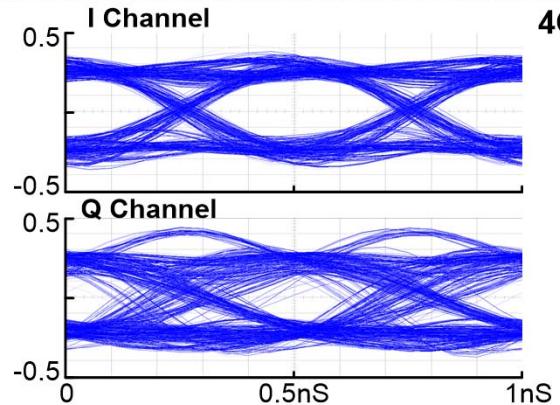
EVM =  $-14\text{dB}$



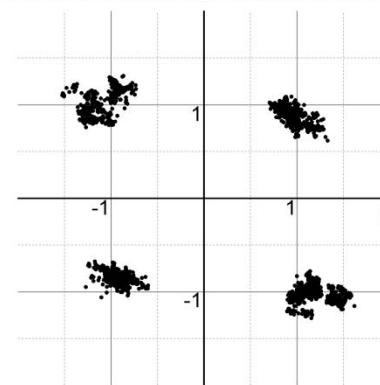
2Gb/s



EVM =  $-12\text{dB}$



4Gb/s



[Y. Wang, ESSCIRC 2015]

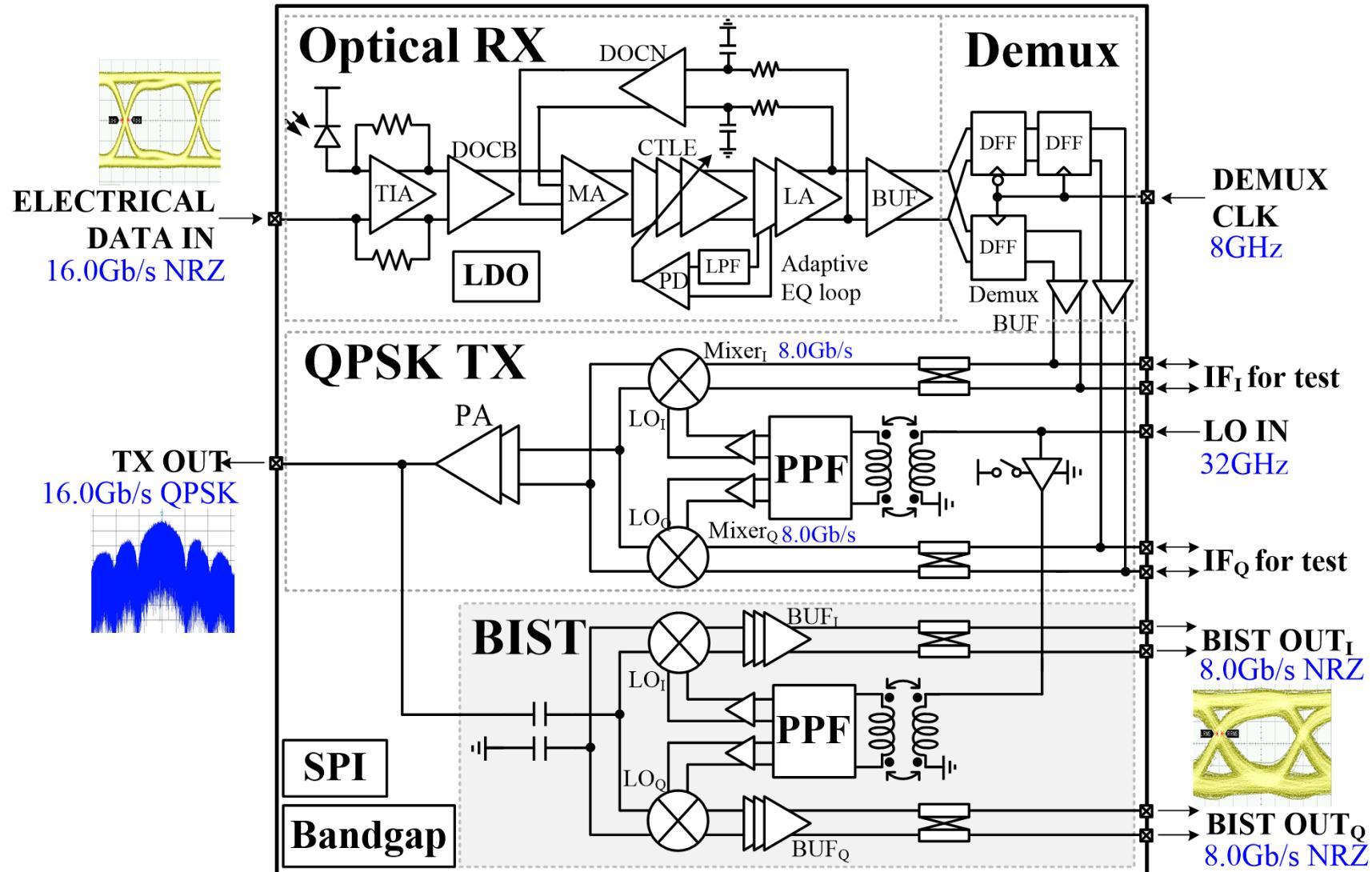


# Outline

- Motivation
- BoF circuit and system design
  - Optical RX with integrated PD
  - Wideband QPSK TX
  - Built-in-self-test
- Measured results
- Summary and future work



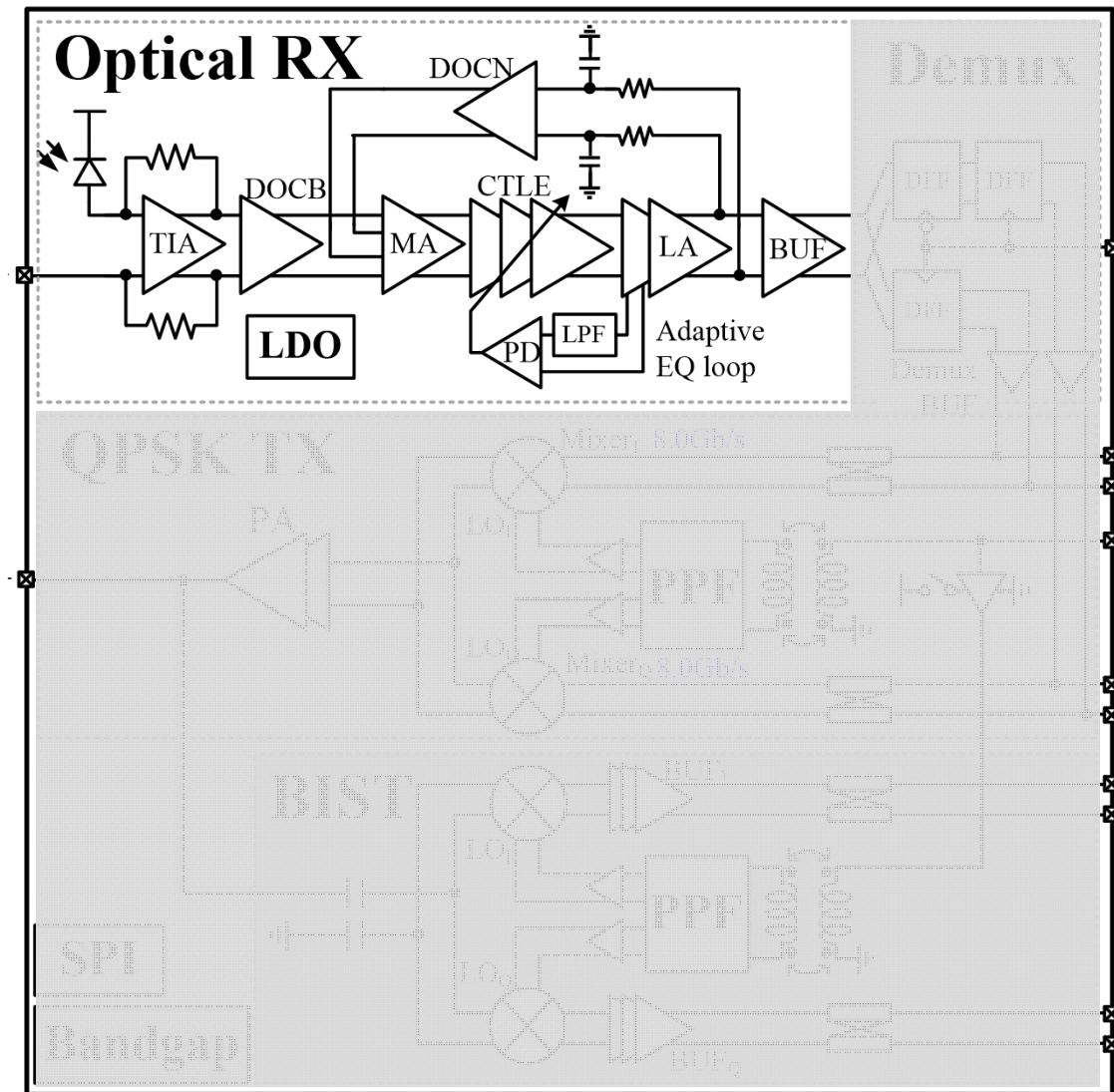
# The BoF System Architecture



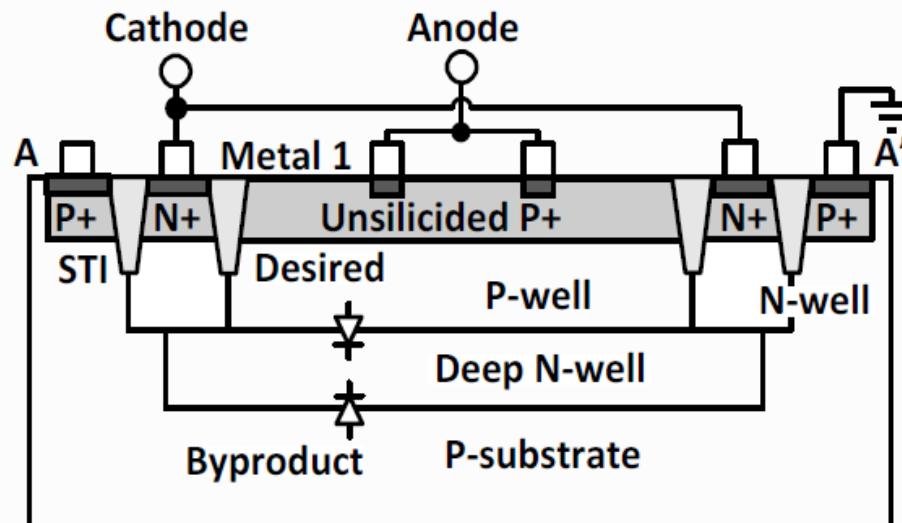
# Optical Receiver

## □ Optical RX

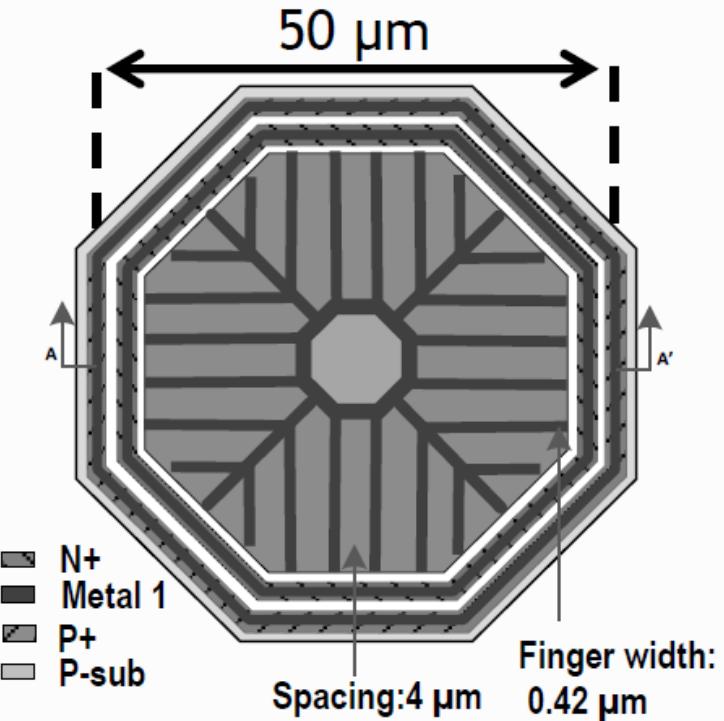
- on-chip PD
- inv-based TIA
- 3-Stage CTLE
- MA/LA



# P-well/Deep N-well PD Structure



Cross-section view

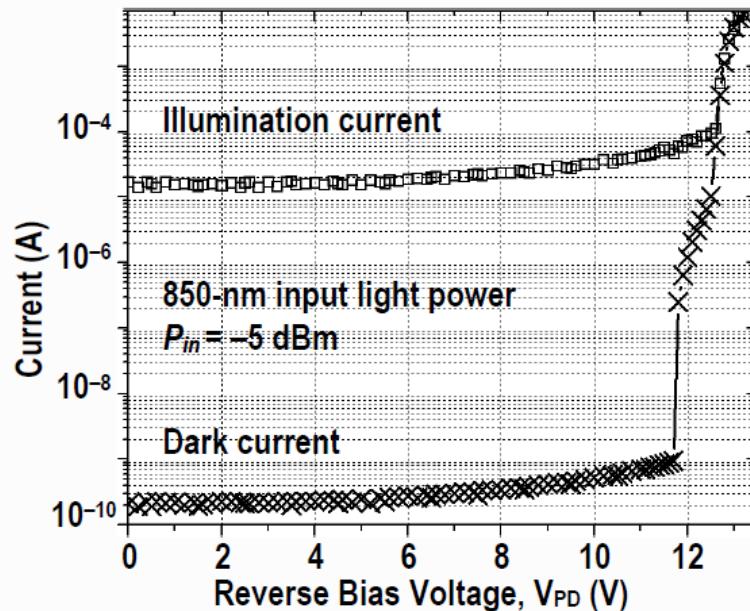


Top-down view

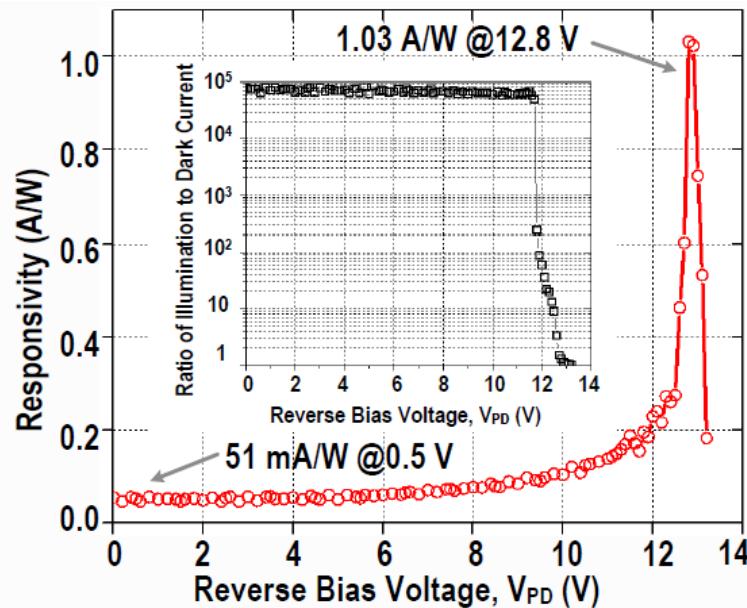
- Eliminate the slow substrate diffusion current
- Deeper junction depth and lighter doping concentration
- Compatible with optical receiver design when operating in the avalanche mode

# P-well/Deep N-well PD Structure

## Measured Responsivity



Measured illumination and dark currents of the PW/DNW PD

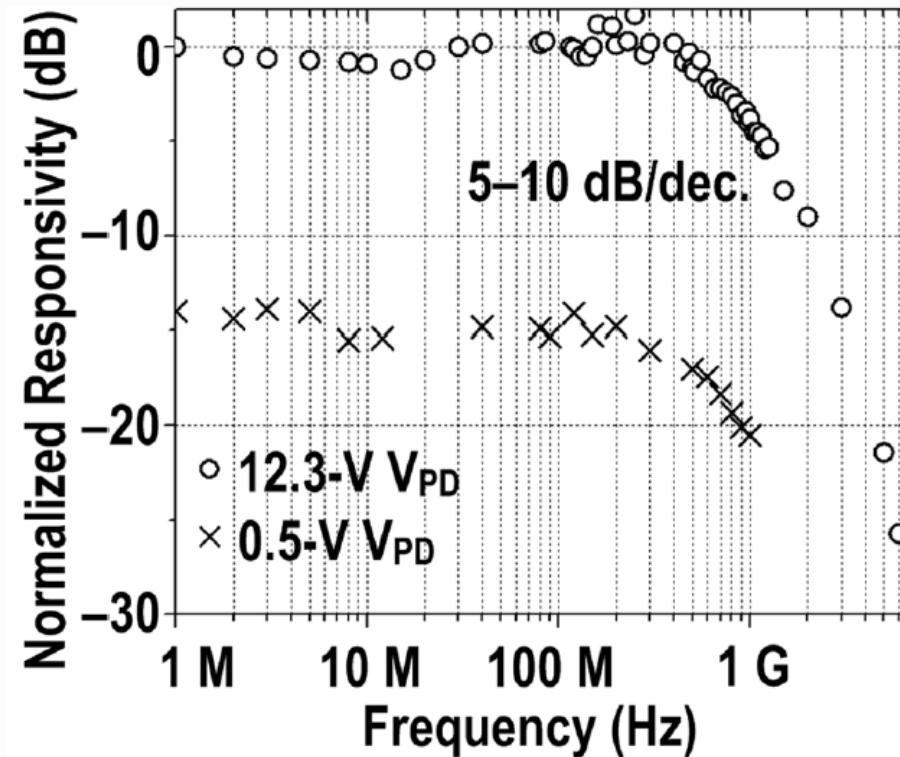


Measured bias dependency of the PW/DNW PD responsivity

- Input light power (-5dBm) is chosen according to sensitivity to work in the linear region
- 51mA/W @ 0.5V; 272mA/W @ 12.3V (optimal); 1.03A/W @ 12.8V (maximum)

# P-well/Deep N-well PD Structure

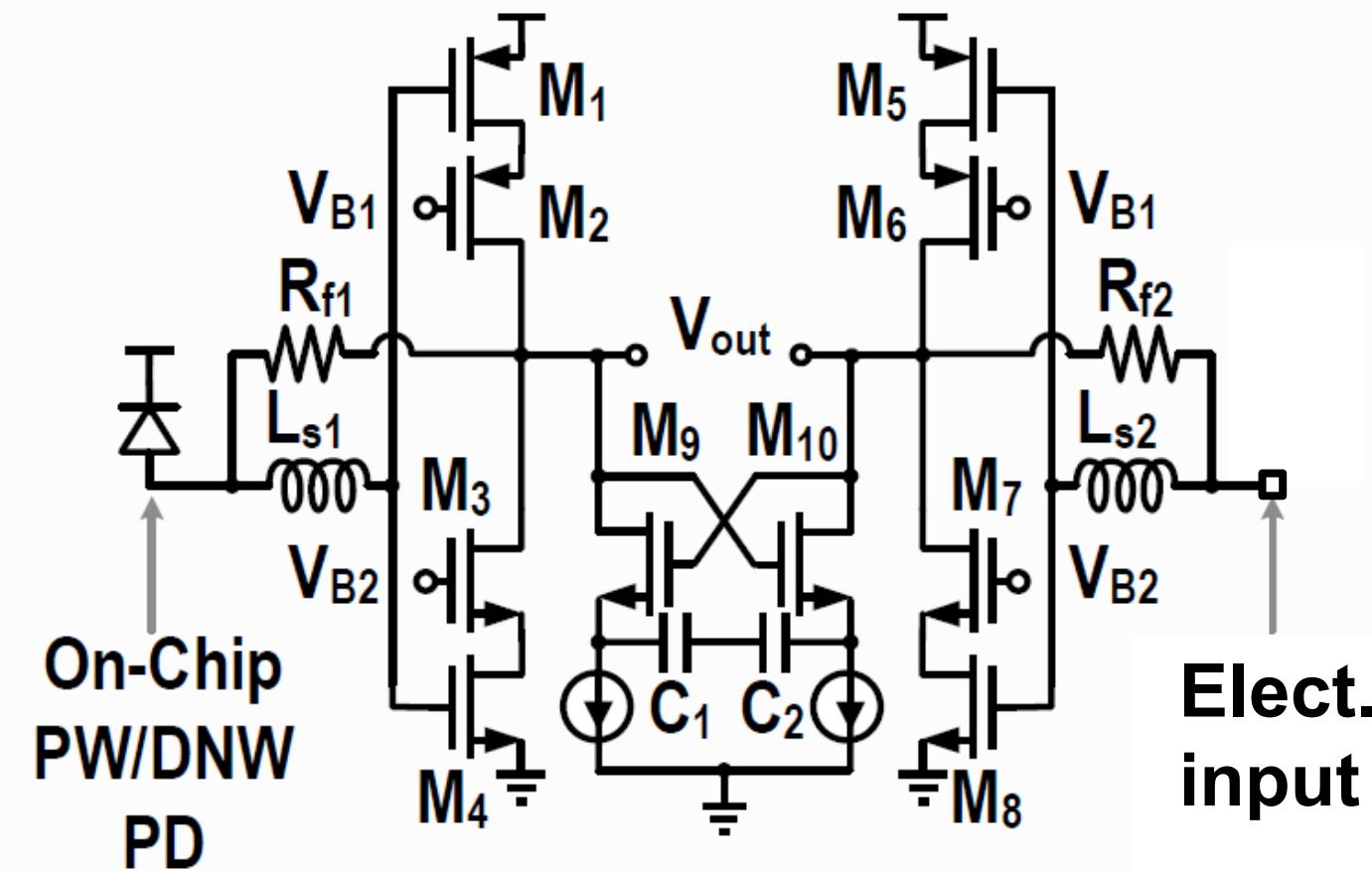
## □ Measured Optical Frequency Response



- Normalized to DC responsivity @ avalanche mode
- Slow roll-off frequency response
- Fitting -3dB bandwidth: 500MHz

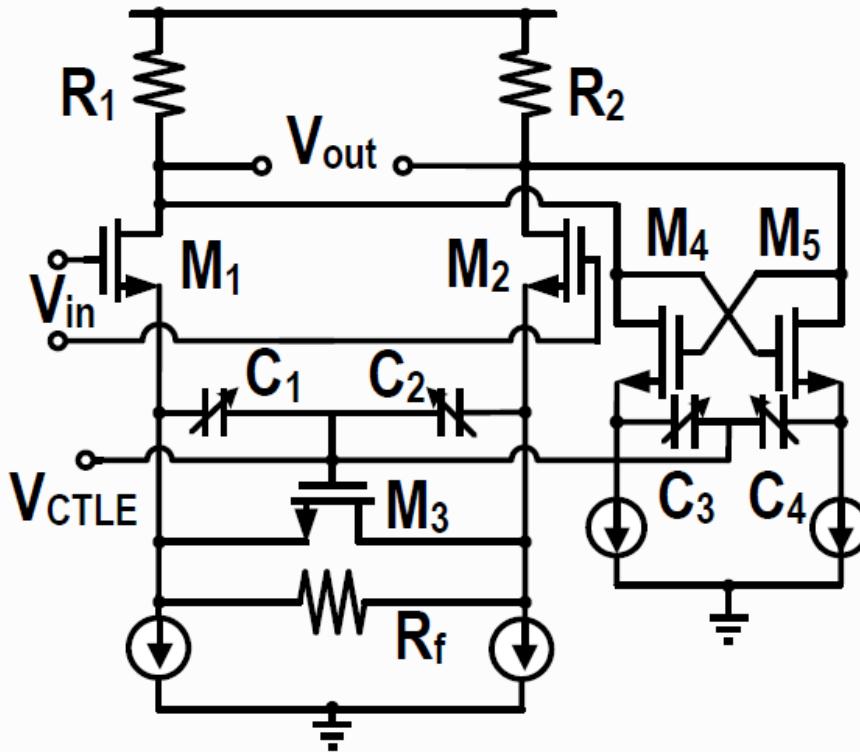


# Inductive Cascode Inverter-Based TIA

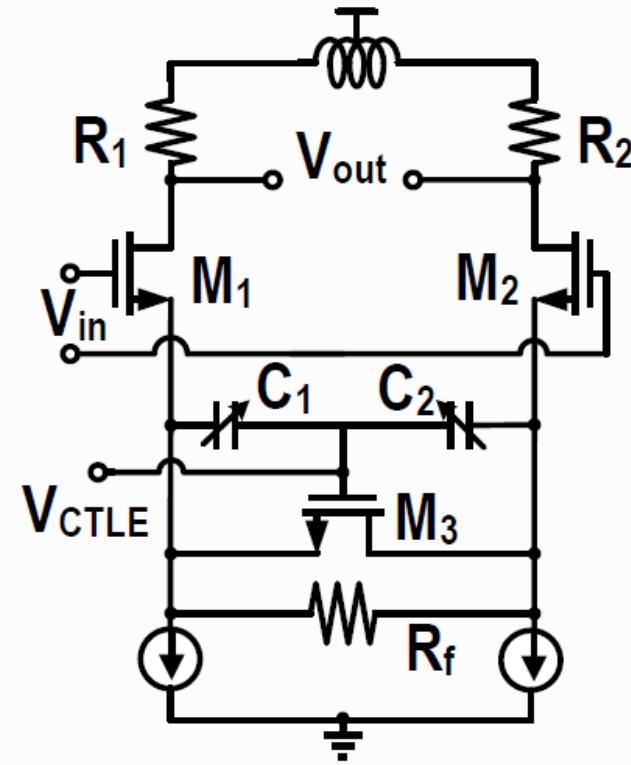


- Given a 480-fF PD capacitance, boost the bandwidth and minimize the input referred noise (IRN)

# Slow Roll-Up 3-Stage Cascaded CTLE



1<sup>st</sup> stage CTLE

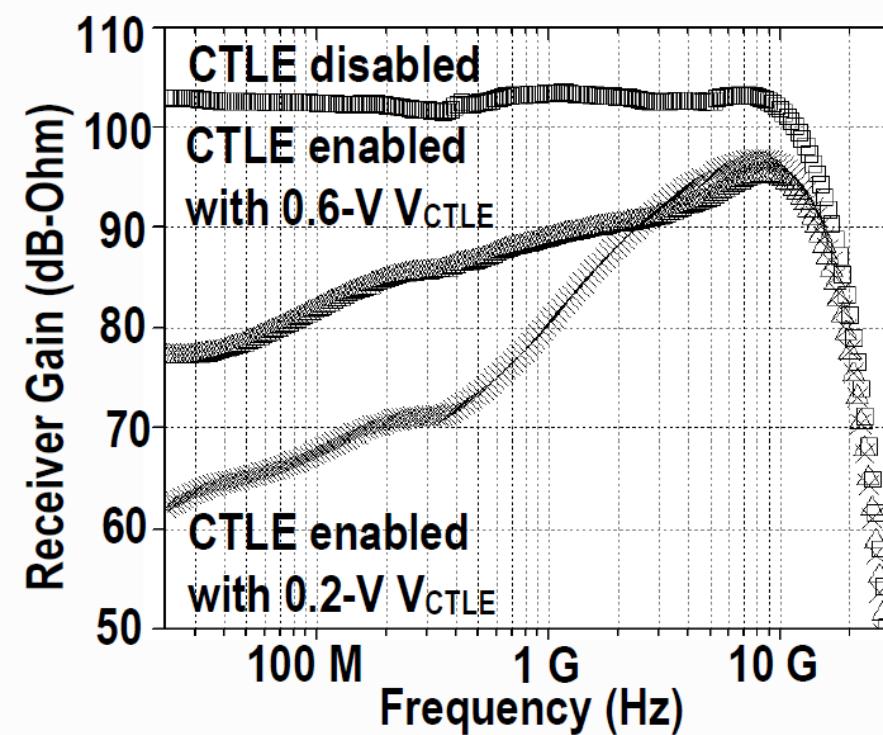


2<sup>nd</sup> and 3<sup>rd</sup> stage CTLE

- By interpolating the poles and zeros, a slow roll-up (5~10dB/decade) response can be achieved

# Slow Roll-Up 3-Stage Cascaded CTLE

- Measured Electrical Frequency Response
  - The gain of optical RX

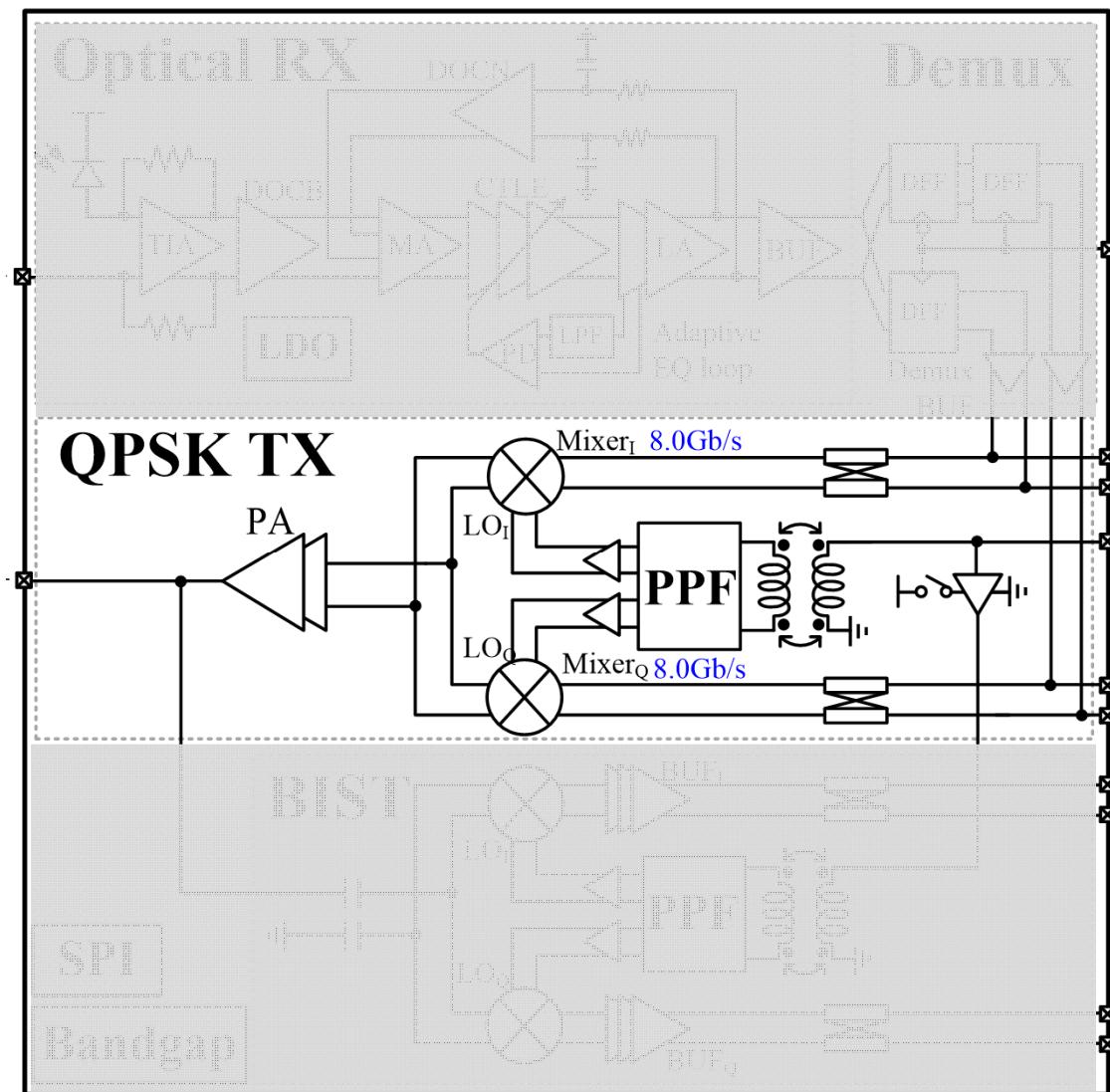


- Tested by direct probing without the on-chip PD
- 33-dB CTLE tuning range

# QPSK Transmitter

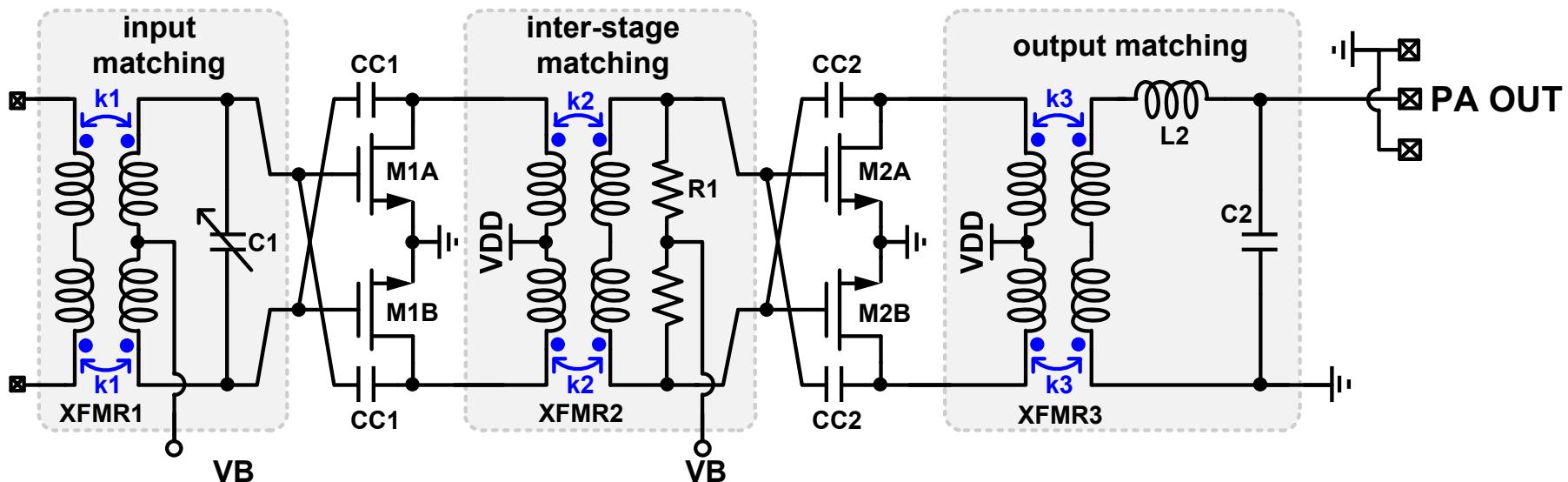
## □ QPSK TX

- 2-stage PPF
- Wideband PA
- Gilbert mixer
- Ripple cancel



# Wideband Power Amplifier

## Wideband Power Amplifier



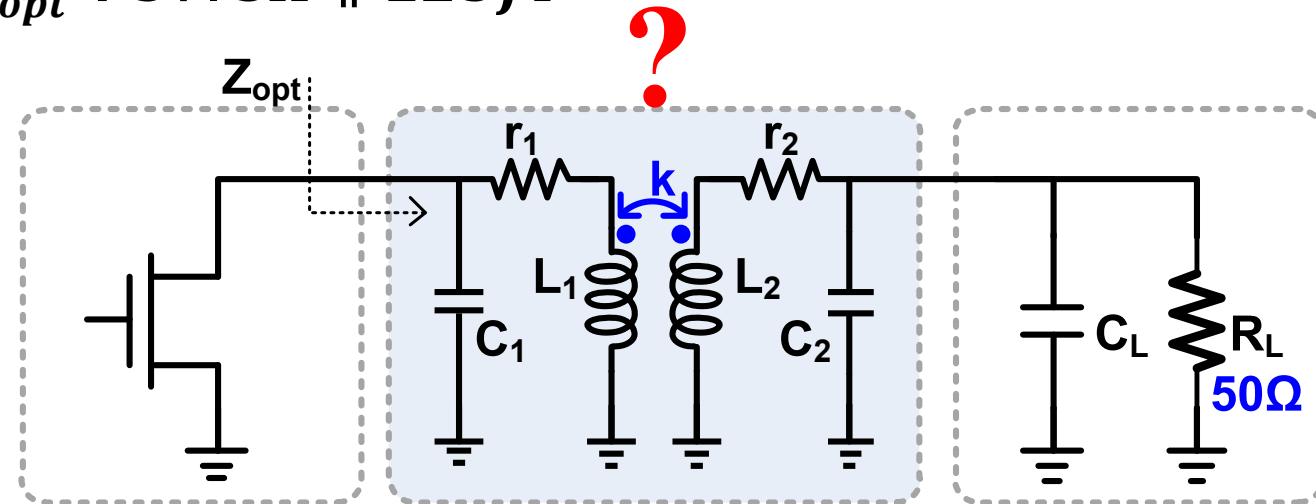
M1A/M1B	CC1	M2A/M2B	CC2
$64\mu\text{m}/60\text{nm}$	$23\text{fF}$	$128\mu\text{m}/60\text{nm}$	$45\text{fF}$

- Two amplification stages with neutralization cap
- Coupled resonator wideband matching network

# Wideband Power Amplifier

## □ Coupled Resonator based Wideband Matching

- Transform the loading to  $Z_{opt}$  obtained from load pull simulation
- $R_L = 50\Omega$ ,  $C_L = 25fF$  (PAD capacitance)
- $Z_{opt}^* : 37.5\Omega \parallel 123fF$

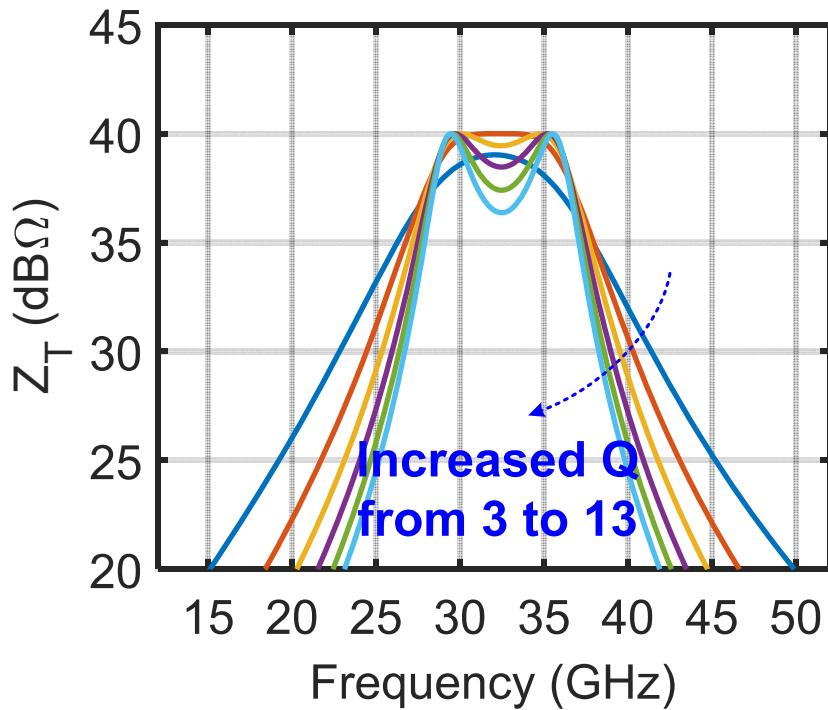
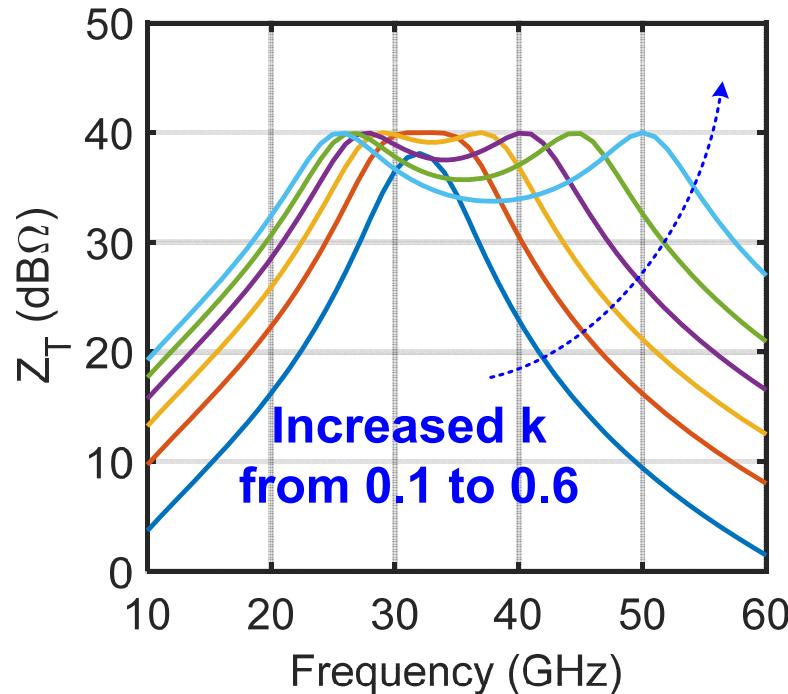


How to achieve wideband impedance transformation with low insertion loss?

# Wideband Power Amplifier

## □ Coupled Resonator based Wideband Matching

### ➤ The effect of $k$ and $Q$



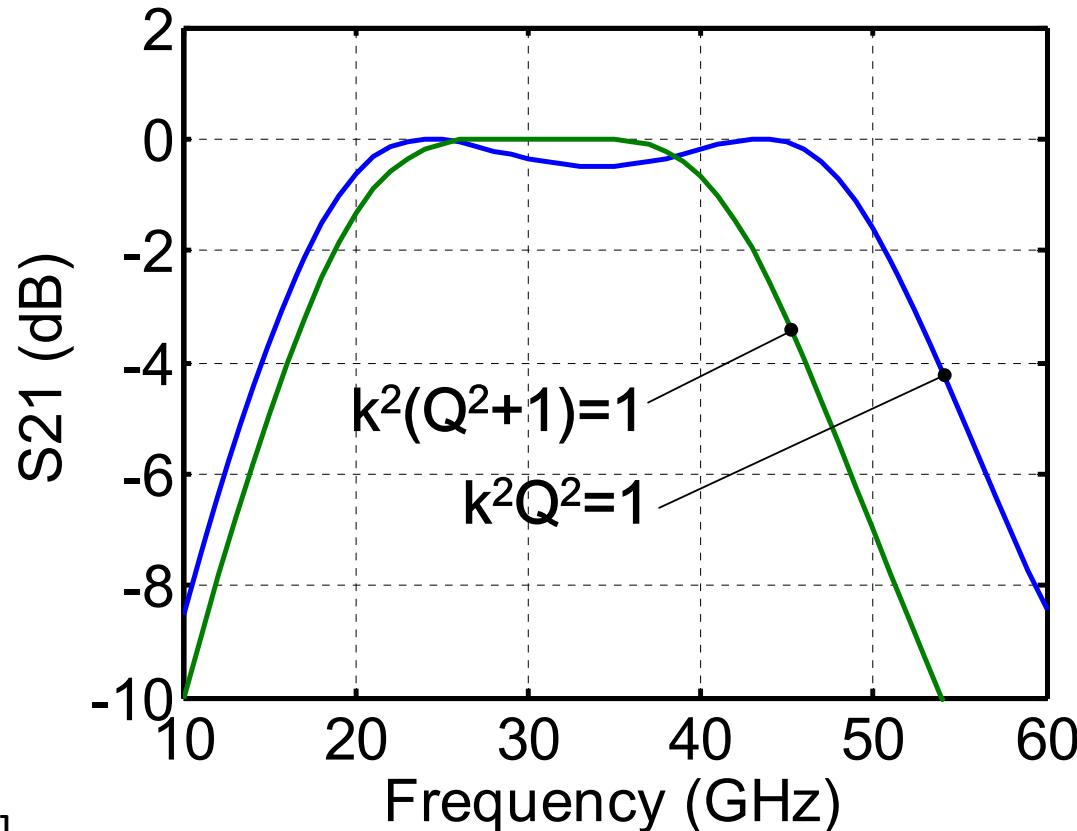
- large  $k$  pushes two poles away
- central freq increases with  $k$
- When is the flattest response?

- large  $Q$  leads to large ripple
- $Q$  has little influence on peak magnitude

# Wideband Power Amplifier

## □ Coupled Resonator based Wideband Matching

- Minimal gain ripple condition:  $k^2(Q^2 + 1) = 1$
- More accurate than previous mentioned  $k^2Q^2 = 1$  [1]

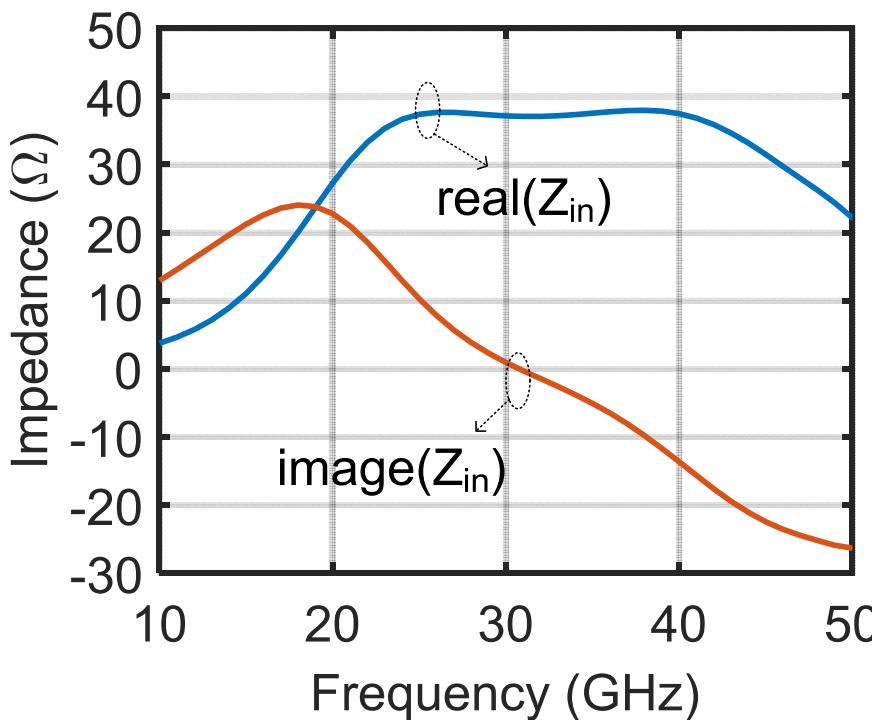


# Wideband Power Amplifier

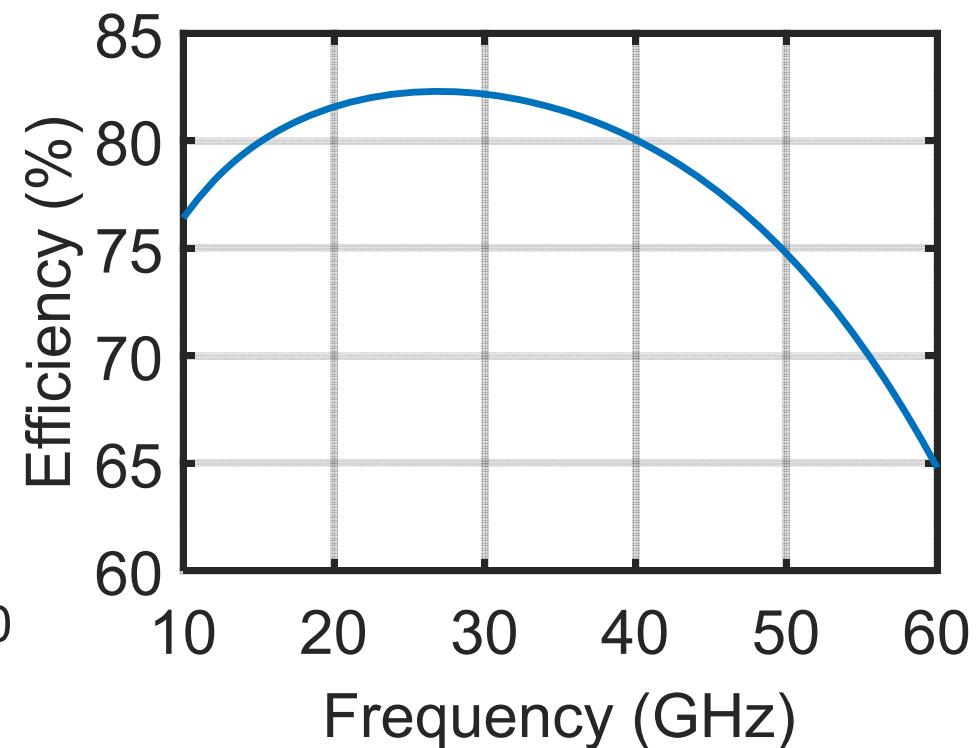
## □ Coupled Resonator based Wideband Matching

➤ The **imp transformation achieved when**  $\frac{L_2}{L_1} = \frac{R_L}{R_s}$

Simulated impedance



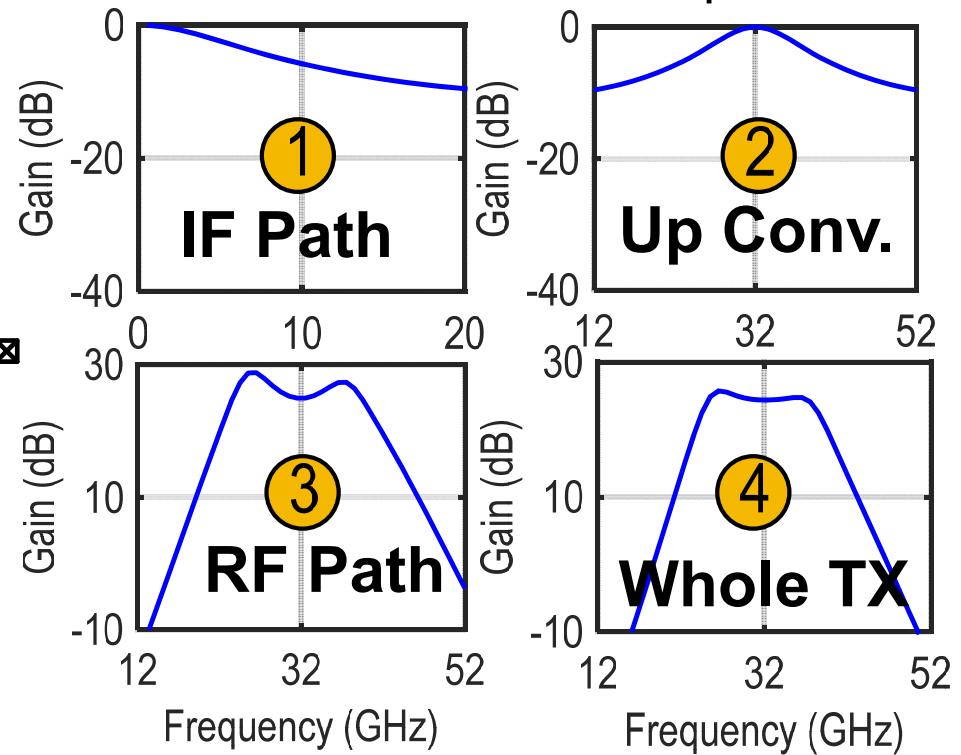
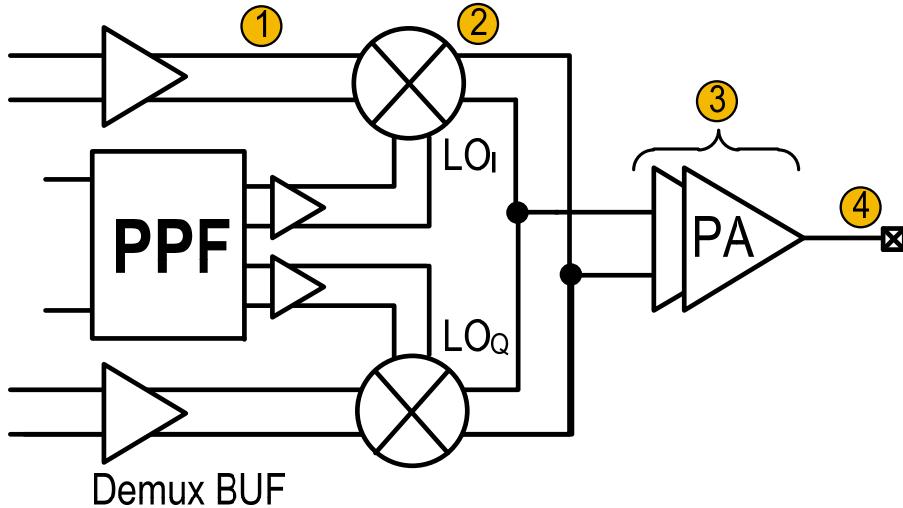
Simulated efficiency



# Wideband Power Amplifier

## □ PA Input/Inter-Stage Matching Network

- Take advantage of gain ripple & bandwidth trade-off
- Use the IF low pass feature to compensate ripple

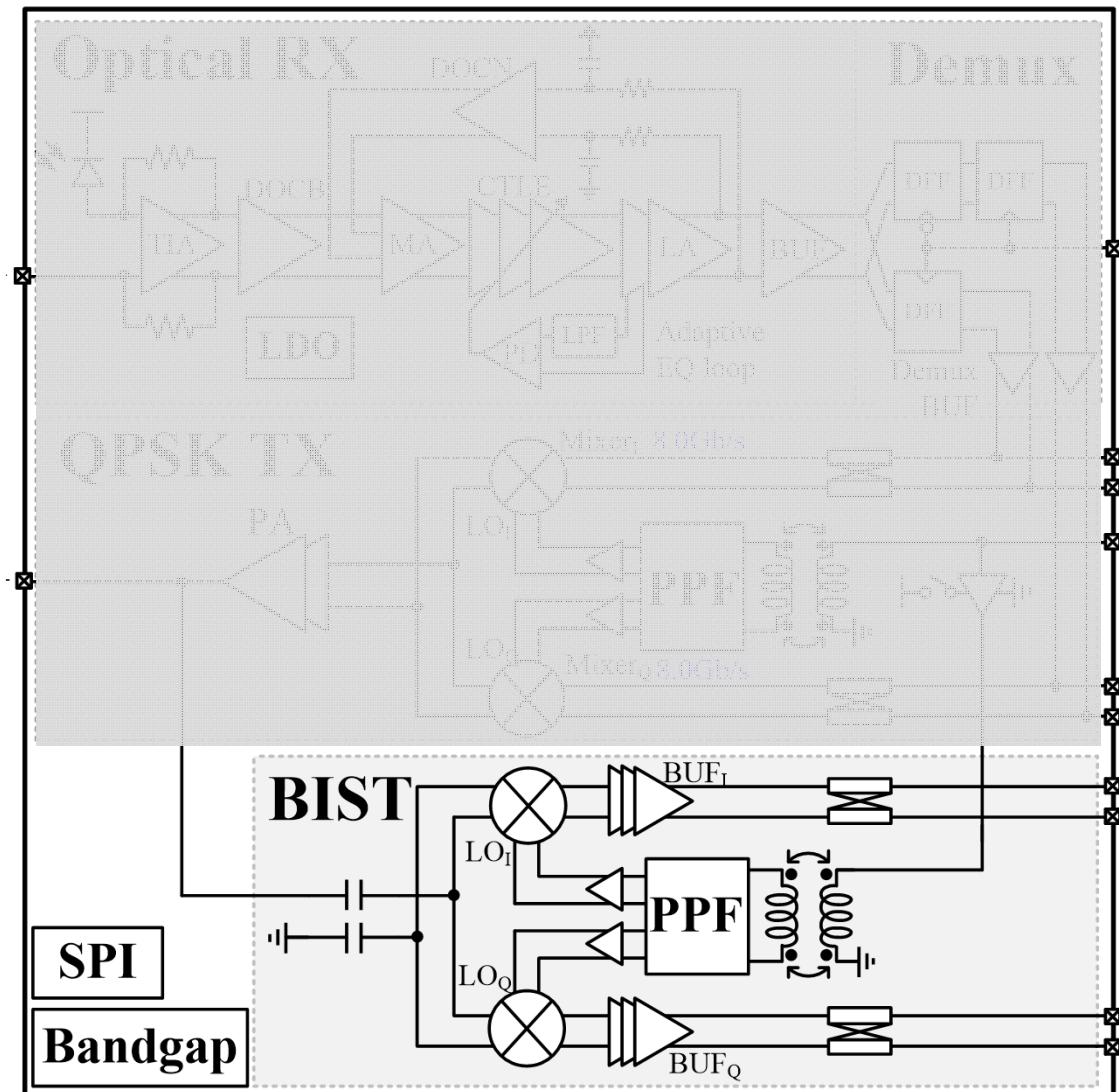


- Simulated gain ripple reduces from 3.8dB to 1.3dB

# Built-In-Self-Test (BIST)

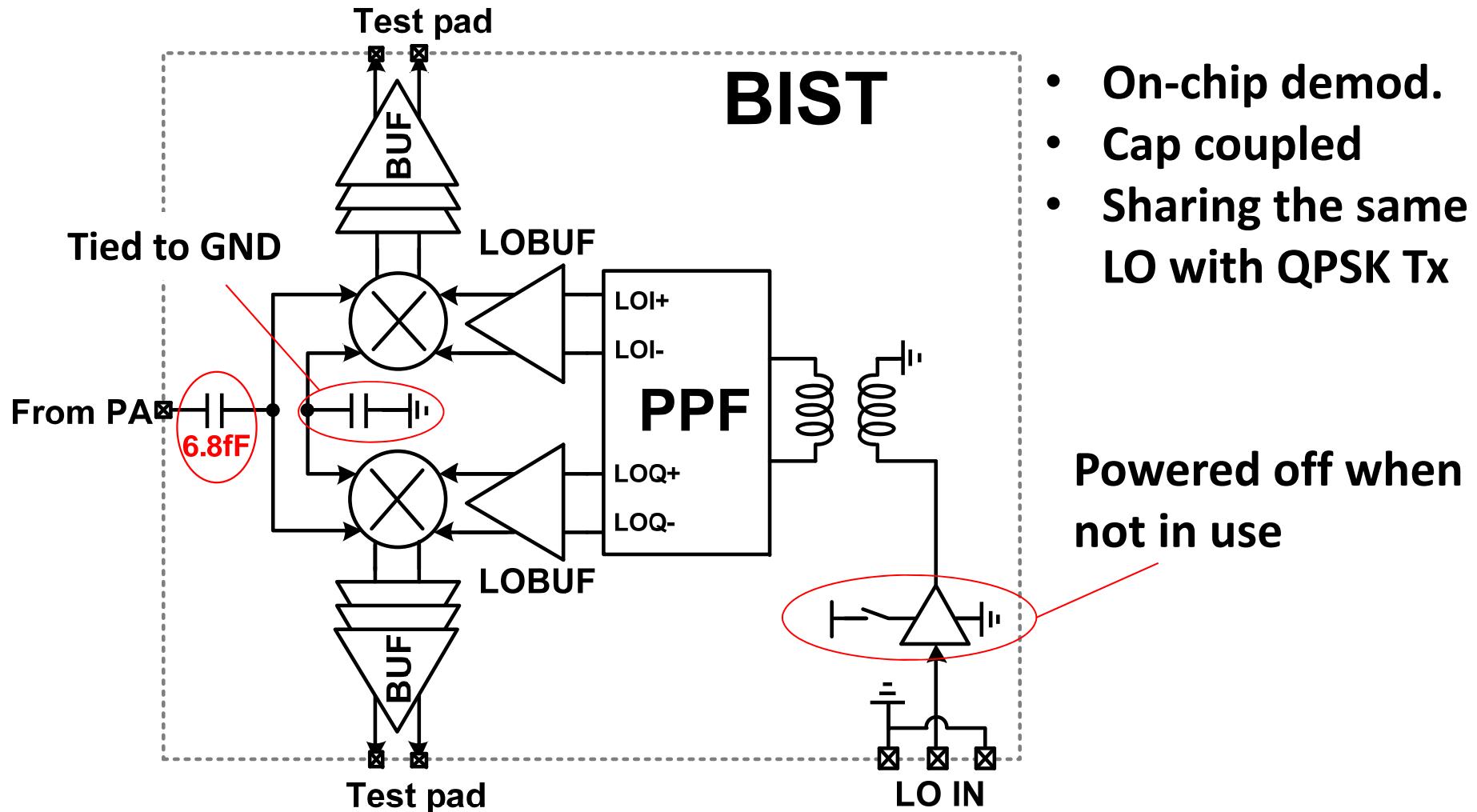
## ❑ BIST

- 2-stage PPF
- Cap coupling
- Current bleeding mixer
- CML driver



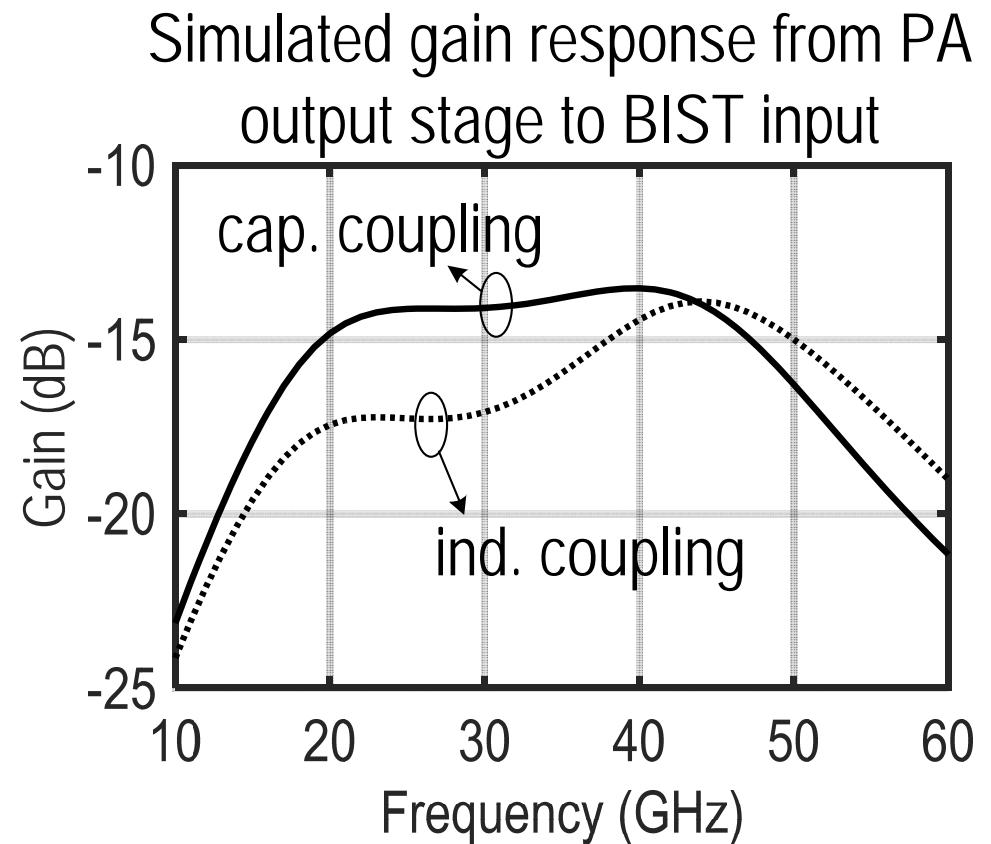
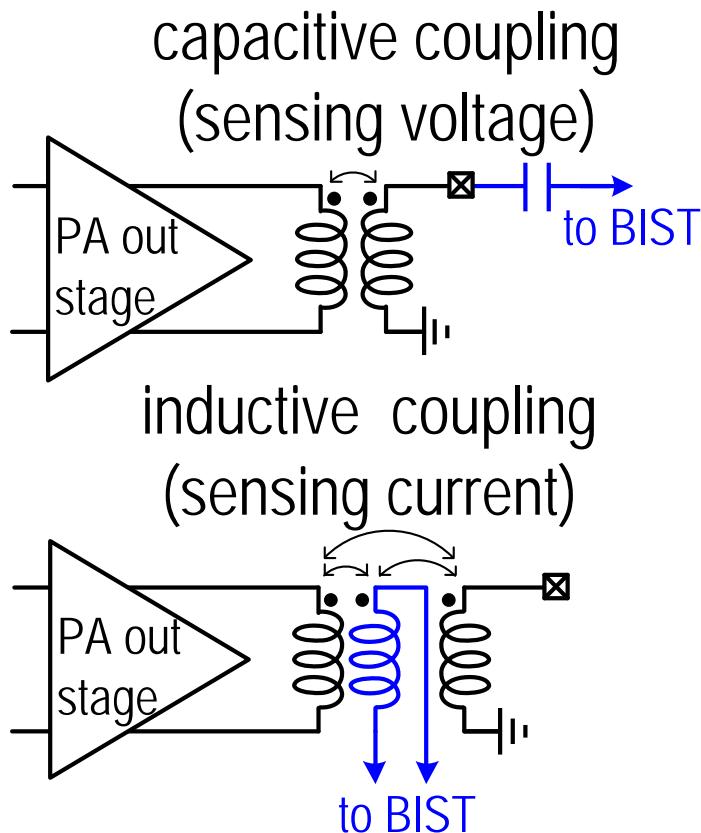
# Built-In-Self-Test

- On Chip Demodulator to support BER test



# Built-In-Self-Test

## □ Cap. Couple Instead of Ind. Couple



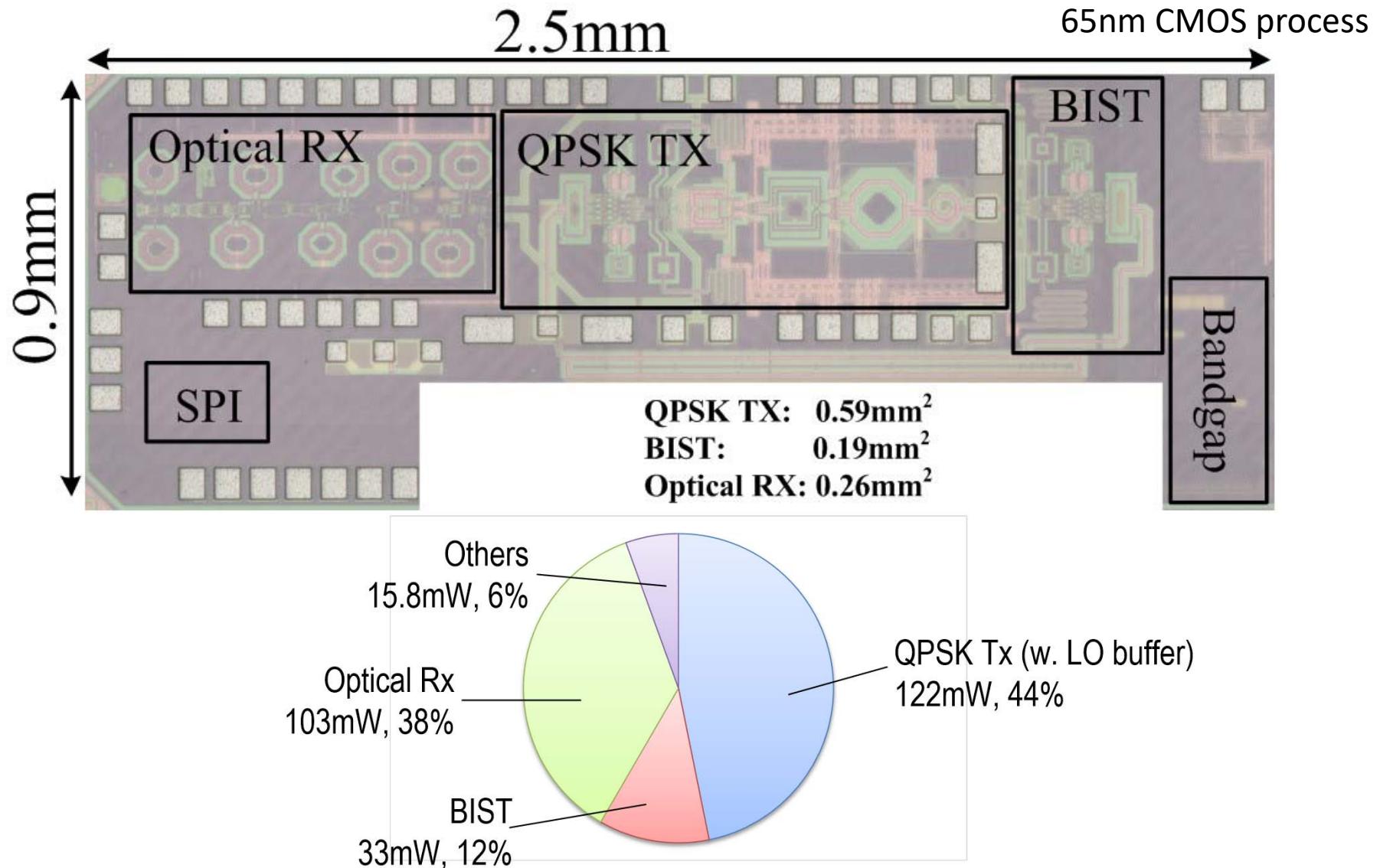
- **Cap coupling maintains the TX's wideband property**



# Outline

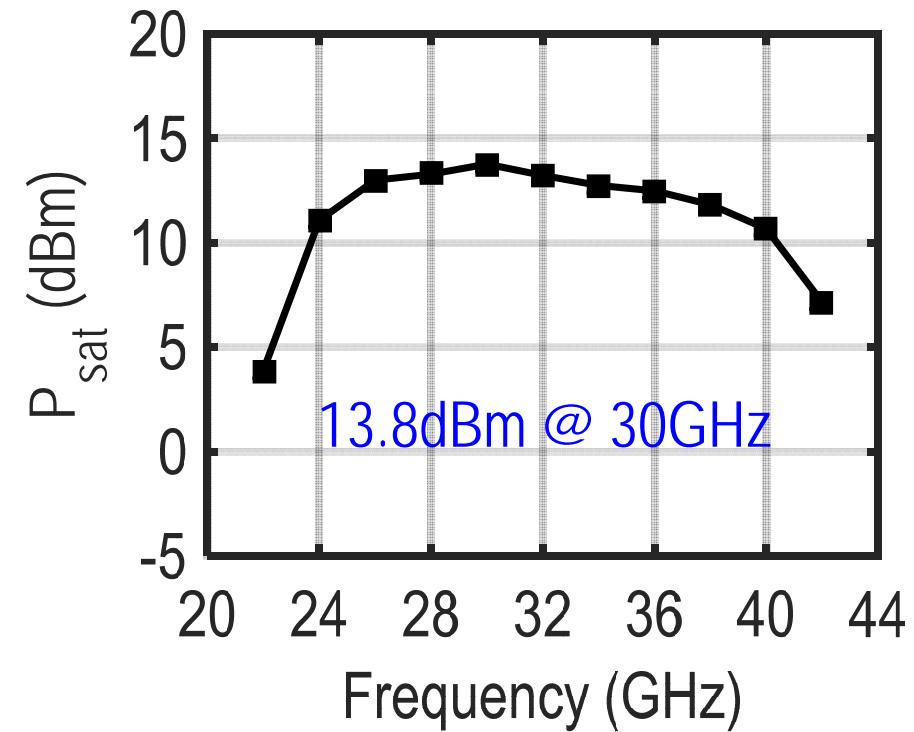
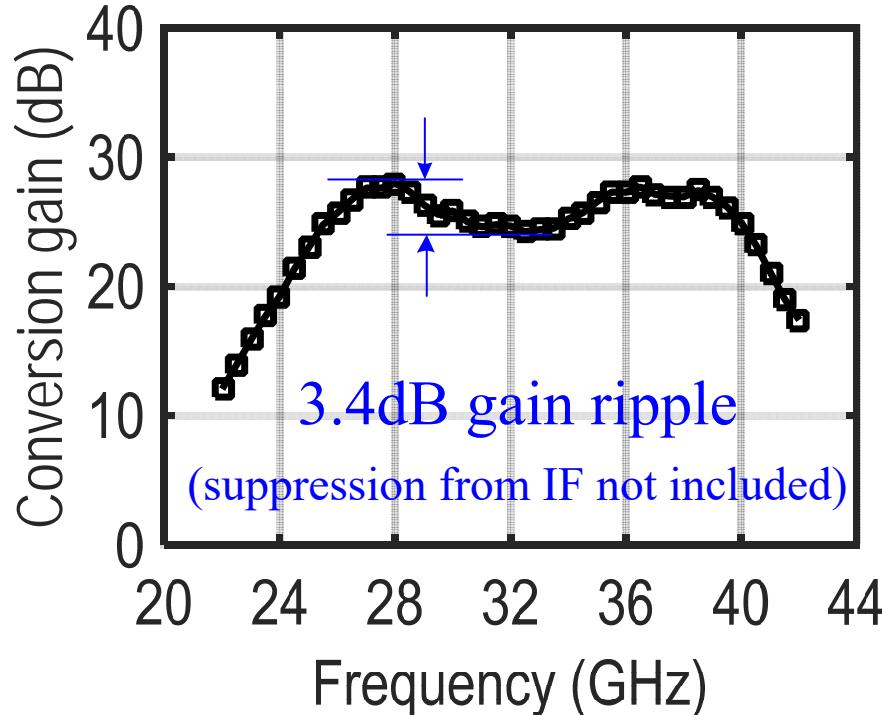
- ❑ Motivation
- ❑ BoF circuit and system design
- ❑ Measured results
- ❑ Summary and future work

# Chip Photo and Power Consumption



# Continuous Wave Measured Results

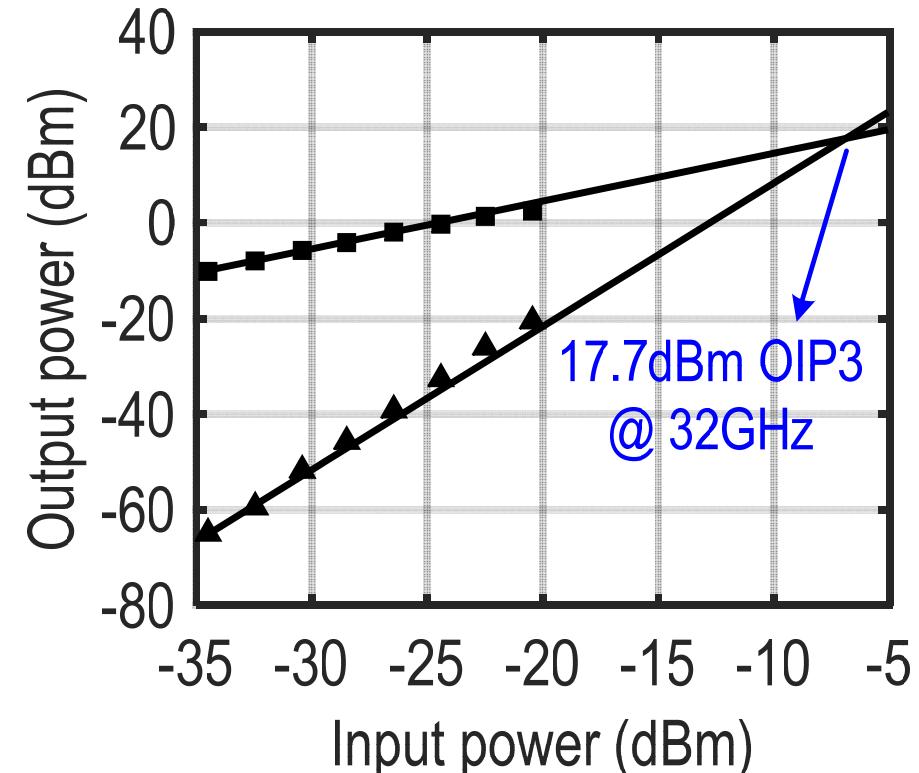
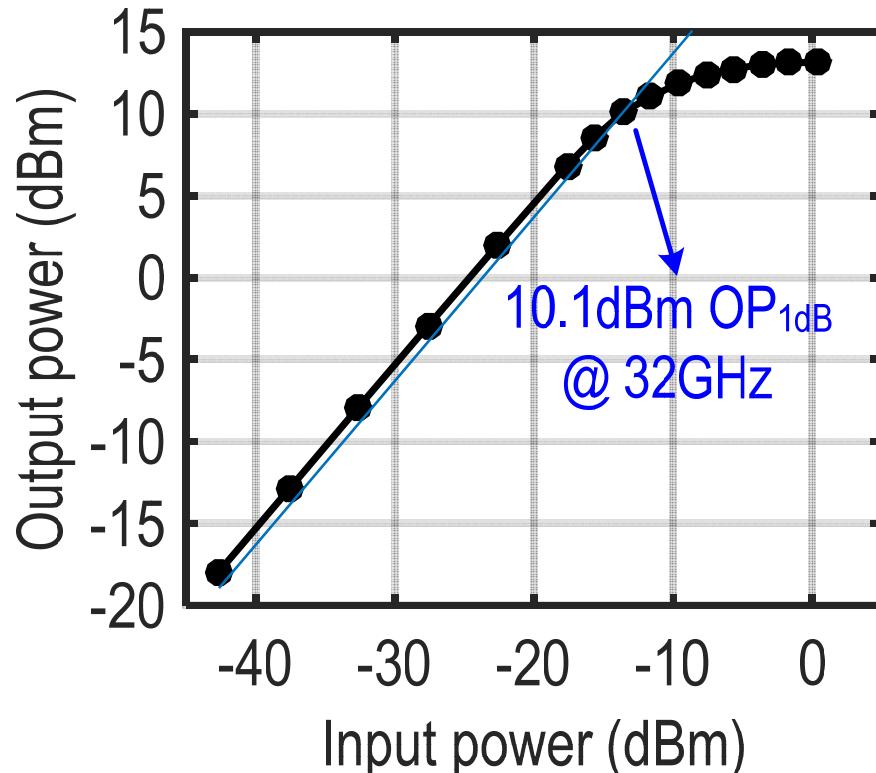
## □ The QPSK TX Conversion Gain



- 24.7dB at 32GHz, >24dB from 25.3 to 40.3GHz
- 3.4dB gain ripple (no include IF path LPF)
- 13.8dBm Psat at 30GHz

# Continuous Wave Measured Results

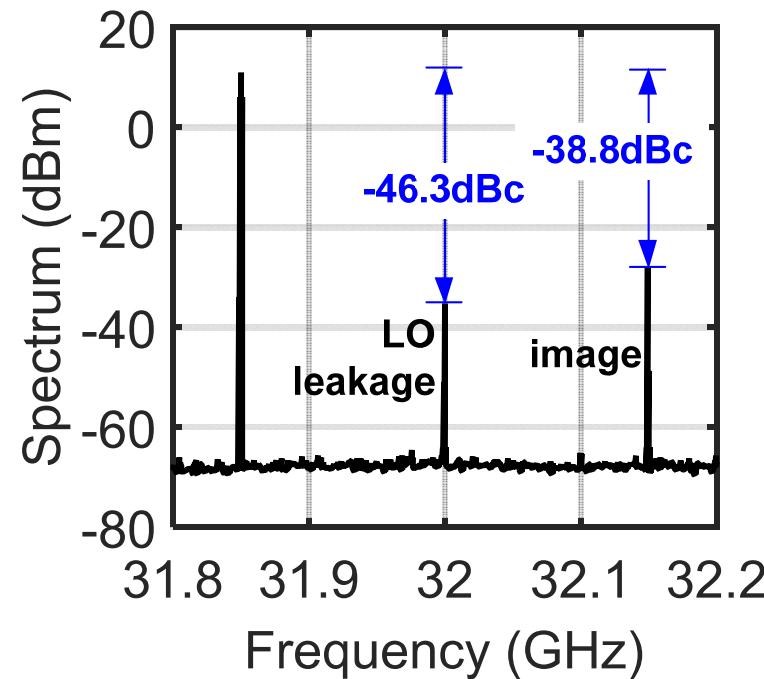
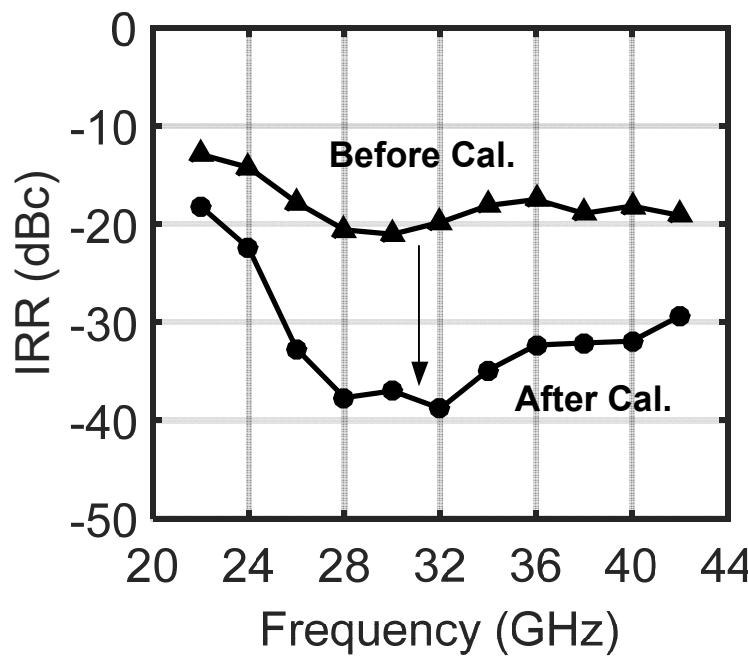
## □ The QPSK TX Large Signal Performance



- Measured 10.1dBm output 1dB point
- Measured 17.7dBm output IP3 point

# Continuous Wave Measured Results

- The QPSK TX Image Rejection Ratio (IRR)
  - Improved by manually tuning the PPF res array

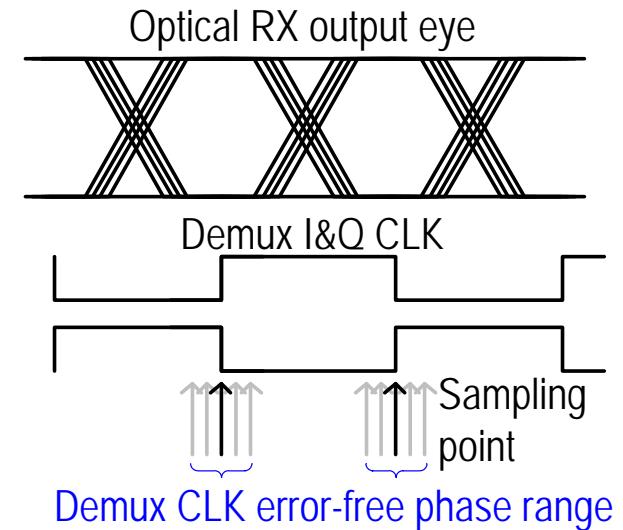
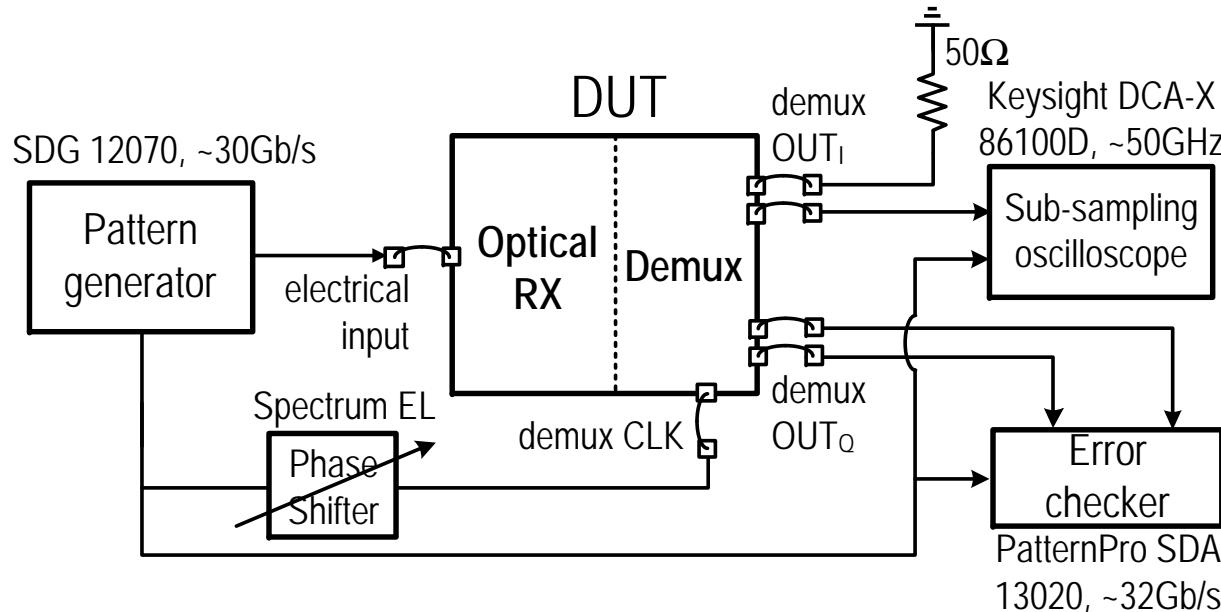


- IRR improved from -19.8dBc to -38.8dBc
- -38.8dBc image rejection and -46.3dBc LO rejection

# Optical RX Measured Results

## □ Measurement Setup

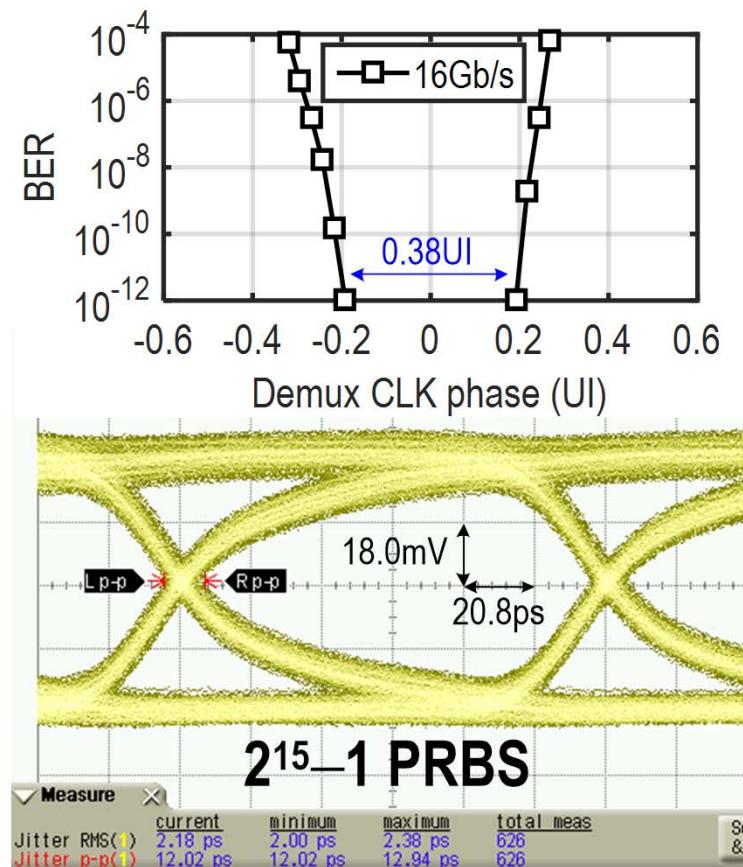
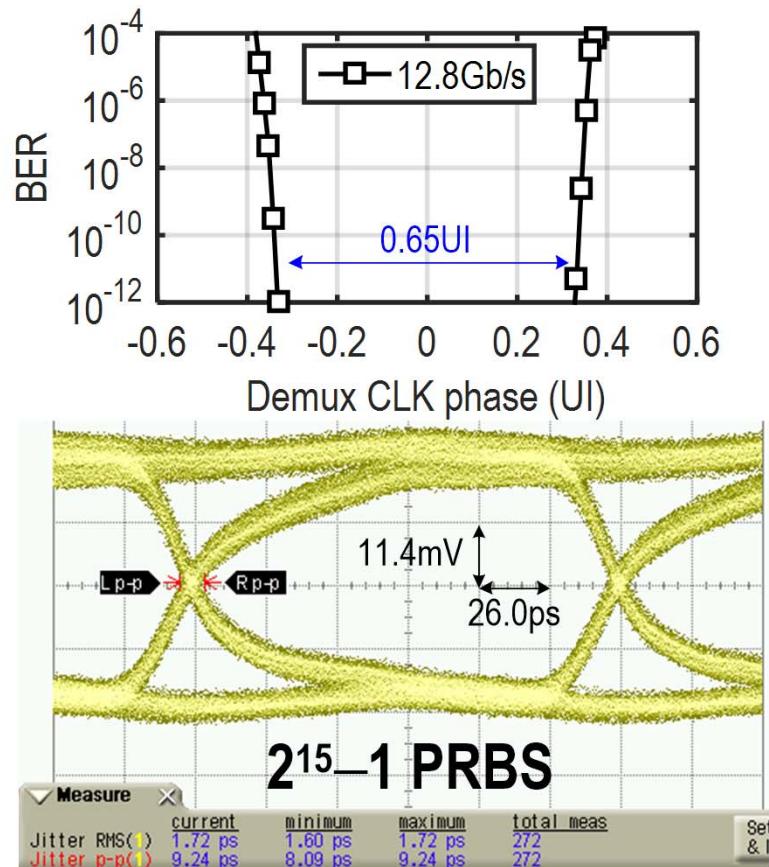
### ➤ External phase shifter



- Manually tuning the phase for optimal sampling
- The error free sampling phase range is measured

# Optical RX Measured Results

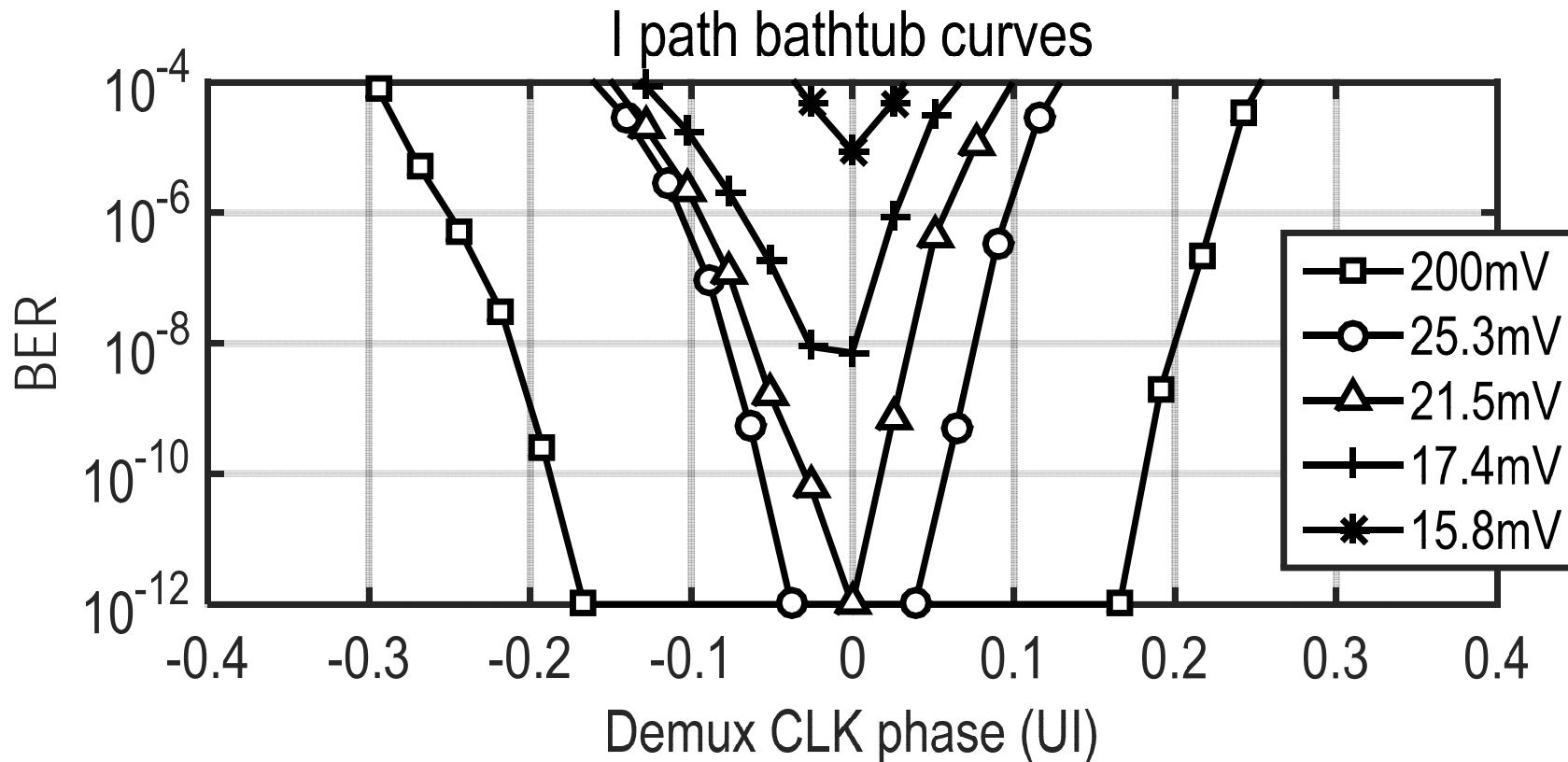
## □ The eye and BER from RX's output



- 0.38UI error free sampling CLK phase range

# Optical RX Measured Results

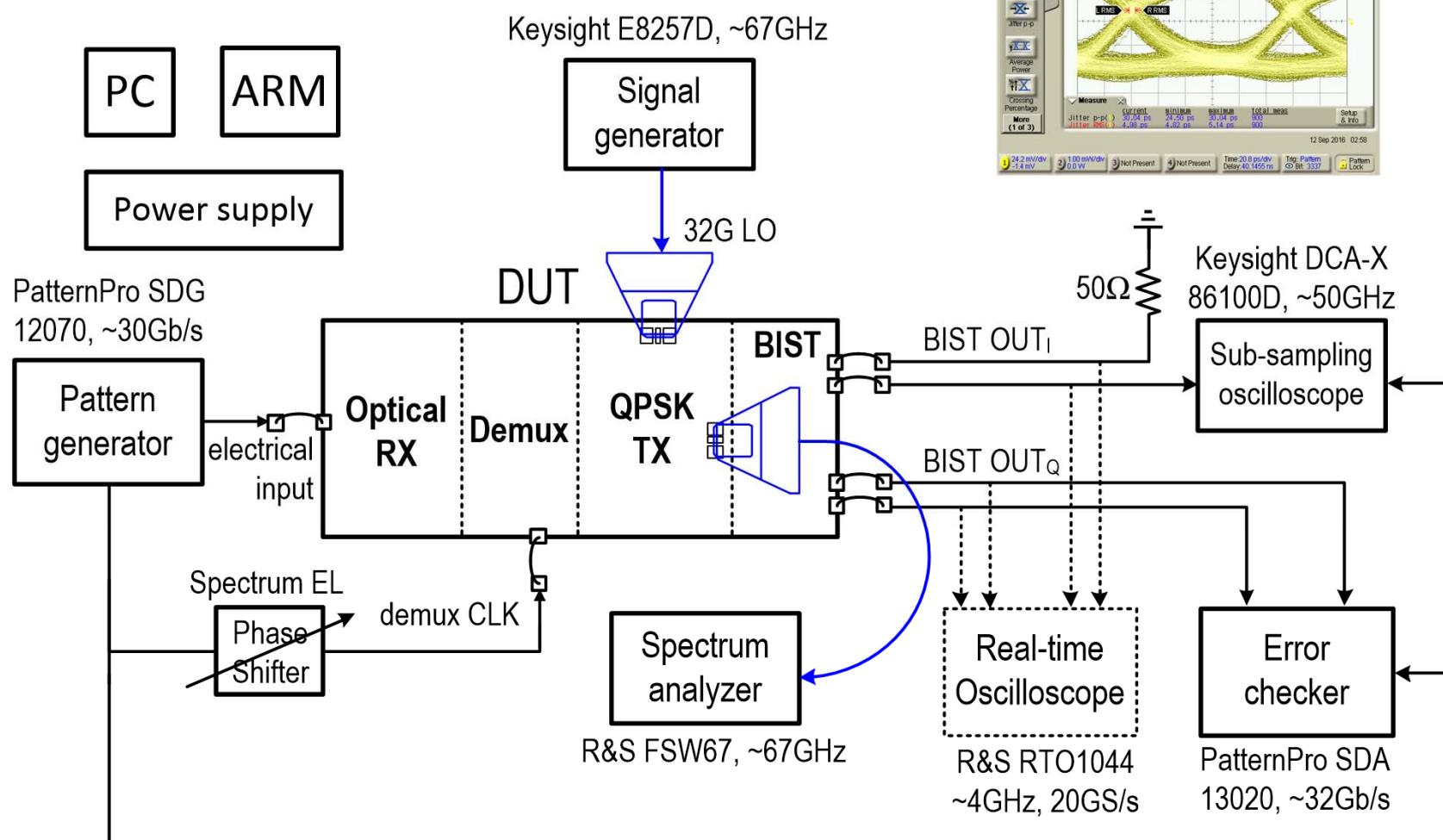
## □ Optical RX sensitivity



- The minimal error free input amplitude is 21.5mV<sub>p-p</sub>
- Converts to 165 $\mu$ A<sub>p-p</sub> using a 80 $\Omega$  input impedance

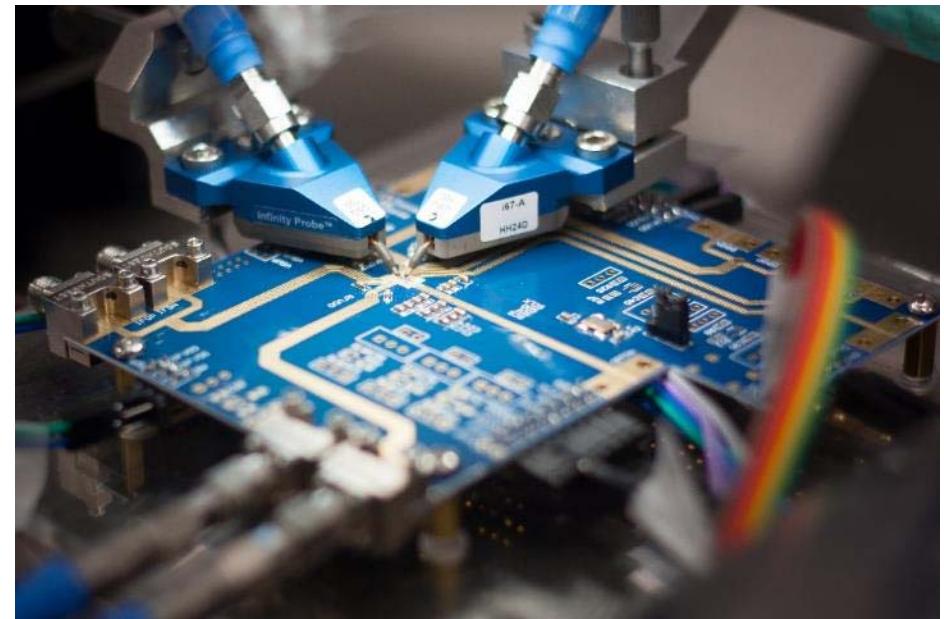
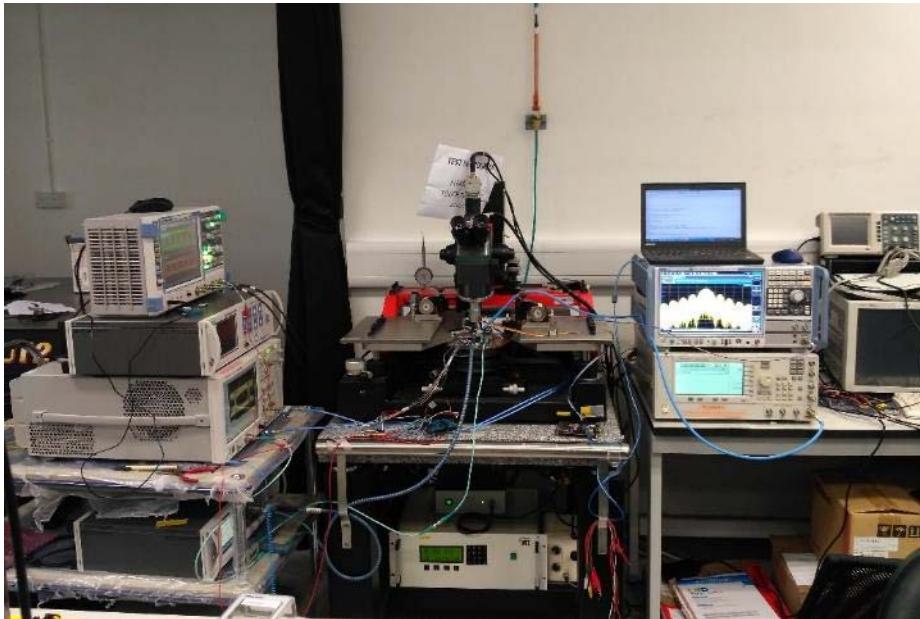
# BoF Data Measured Results

## Measurement Setup



# BoF Data Measured Results

## □ Measurement Setup

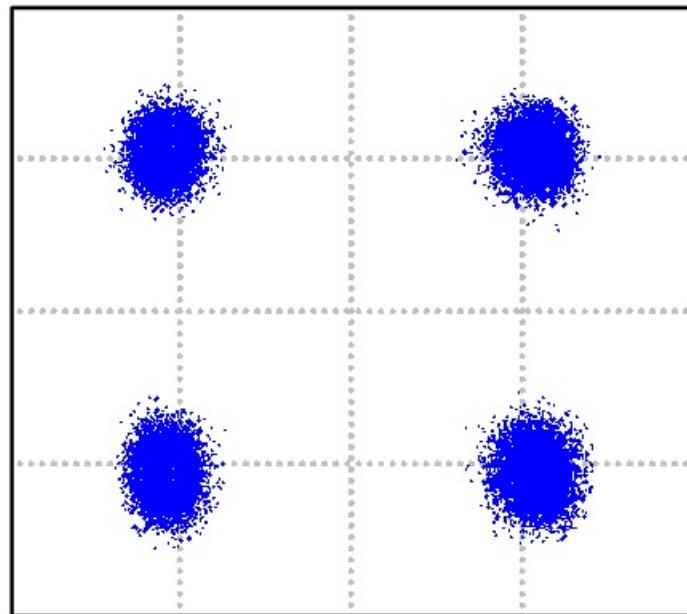


- **On-chip probing for high speed signals**
- **BERT system, signal gen, subsampling osc, real time osc**

# BoF Data Measured Results

## □ The EVM of TX's output

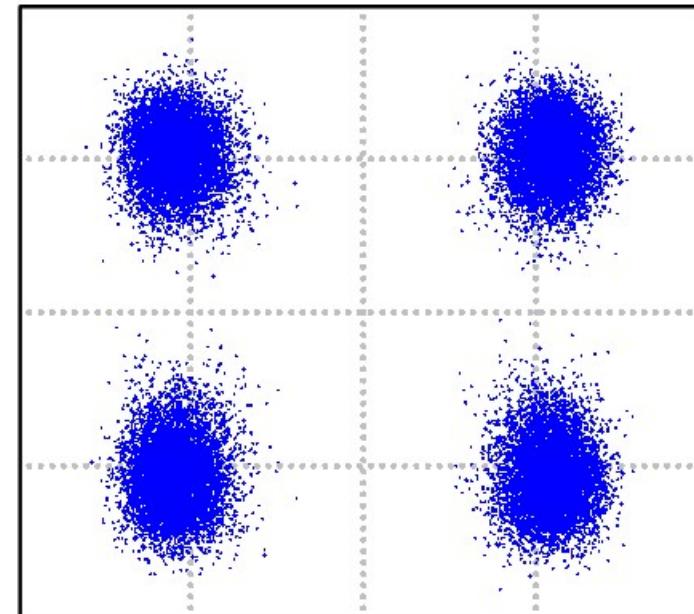
6.4Gb/s QPSK



EVM=-18.8dB

(32767 symbols)

12.8Gb/s QPSK



EVM=-15.6dB

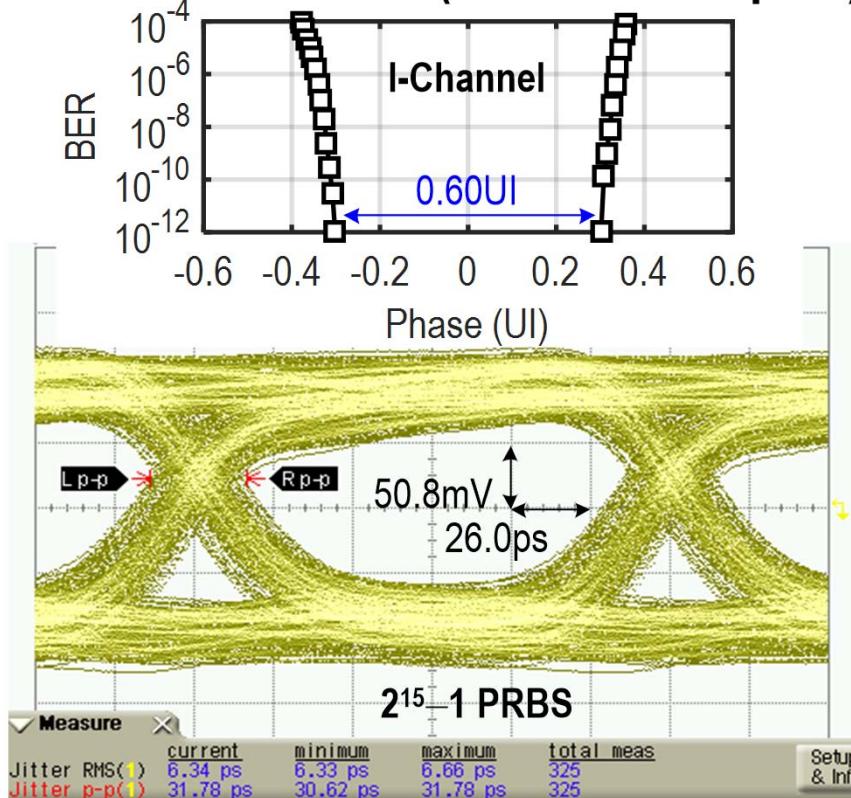
(32767 symbols)

- Using  $2^{15}-1$  PRBS pattern, 32767 symbols collected
- Degraded due to insufficient OSC's BW

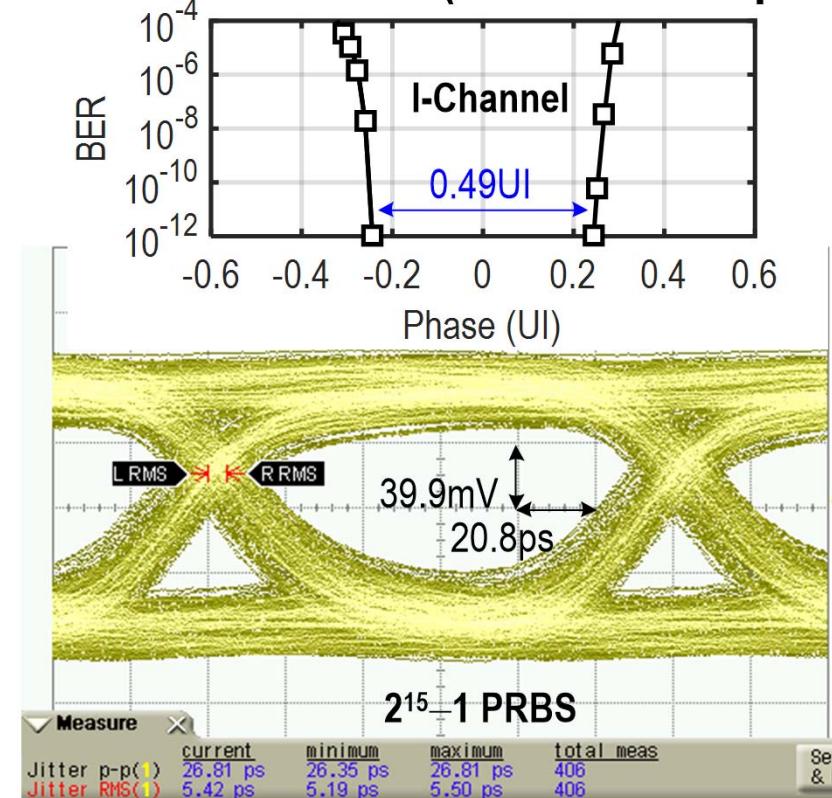
# BoF Data Measured Results

## □ The eye and BER from BIST's output

12.8Gb/s data rate (6.4Gb/s each path)



16.0Gb/s data rate (8.0Gb/s each path)



- 0.49UI error free range at 16.0Gb/s data rate



# Performance Summary and Comparison

	This work	ISSCC15 [2]	ISSCC14 [3]	ISSCC13 [4]
Process	65nm CMOS	65nm CMOS	65nm CMOS	40nm LP CMOS
Freq. (GHz)	32	60	60	60
DR (Gb/s) (Mod.)	16.0 (QPSK)	7 (16QAM)	14.08/28.16 (QPSK/16QAM)	3.5/7 (QPSK/16QAM)
TX EVM (dB)	-15.6* (TX-to-BIST)	-21	-20.1/-20.0	-14.8/-15.2 (TX-to-RX)
TX P <sub>DC</sub> (mW) Excl. LO	138	174	186	181***
TX Eff. (pJ/b)	8.6	24.9	13.2/6.6	78.4/39.2
CG (dB)	24.7	N/A	15	31
P <sub>-1dB</sub> (dBm)	10.1	6.3	10.3**	10.8***
BIST	Yes	No	No	No
BER	<10 <sup>-12</sup> (TX-to-BIST)	N/A	N/A	<10 <sup>-4</sup> (TX-to-RX at 3.6m)
TX frac. BW (%)	46	N/A	N/A	10
TX Area (mm <sup>2</sup> )	0.59	0.79	1.03	1.5*

\*EVM at 12.8Gb/s

\*\*Saturated output power

\*\*\*Each TX path



## Summary

- Wideband 32GHz QPSK with integrated optical RX and BIST is presented
- Coupled resonator based wideband matching
  - Ripple cancellation using IF intrinsic LPF
- An error-free 16.0Gb/s data rate is achieved at 8.6pJ/b bit efficiency